

**A TRUST REGION AGGRESSIVE SPACE MAPPING TECHNIQUE
FOR EM OPTIMIZATION**

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SOS-98-15-V

June 1998

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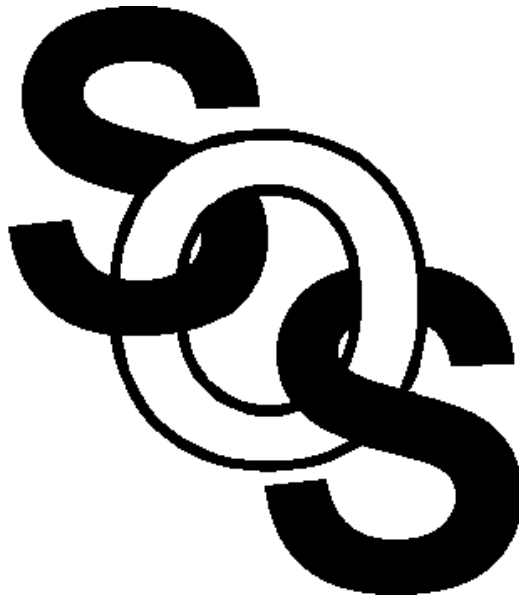
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A TRUST REGION AGGRESSIVE SPACE MAPPING ALGORITHM FOR EM OPTIMIZATION

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presented at

1998 IEEE MTT-S Int. Microwave Symposium, Baltimore, MD, June 1998



Basic Concepts

it is assumed that the circuit under consideration can be simulated using two models: a fine model and a coarse model

the fine model is accurate but is computationally intensive

\mathbf{x}_{em} is the vector of fine model parameters

the coarse model is a fast model but it is less accurate than the fine model

\mathbf{x}_{os} is the vector of coarse model parameters

Space Mapping (SM) aims at avoiding the computationally intensive direct optimization of the fine model

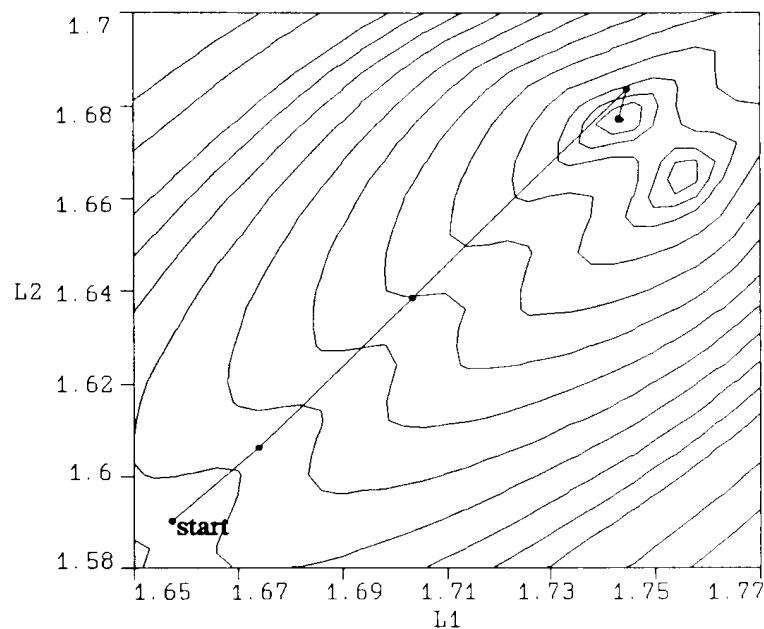


Comments on the Parameter Extraction Step

the parameter extraction step aims at finding a point \mathbf{x}_{os} whose coarse model responses match corresponding fine model responses

the extracted coarse model parameters may be nonunique

nonuniqueness of the parameter extraction step hinders the convergence of the Aggressive Space Mapping technique



contours of the parameter extraction objective function
(*Bandler et al., 1996*)



The Multi-Point Parameter Extraction

to overcome problems caused by nonuniqueness of the parameter extraction step a multi-point parameter extraction step was suggested (*Bandler et al., 1996*)

the step aims at matching not only the response but also the first-order derivative of the two models

the extracted coarse model point \mathbf{x}_{os} is obtained by solving

$$\underset{\mathbf{x}_{os}}{\text{minimize}} \left\| \mathbf{R}_{os}(\mathbf{x}_{os} + \Delta\mathbf{x}) - \mathbf{R}_{em}(\mathbf{x}_{em} + \Delta\mathbf{x}) \right\|$$

simultaneously for a set of perturbations $\mathbf{D}\mathbf{x}$

this step is more likely to improve the uniqueness of the parameter extraction step



Questions Regarding the Original Multi-Point Extraction

the original multi-point parameter extraction (*Bandler et al., 1996*) suffers from two main drawbacks

the first drawback is that the fine model points were arbitrarily selected

the second drawback is the assumption that a perturbation of $\Delta\mathbf{x}$ in the fine model space corresponds to an equal perturbation in the coarse model space

in the i th iteration, the most recent information about the mapping between the two spaces is given by the matrix $\mathbf{B}^{(i)}$, which should be integrated with the multi-point parameter extraction step



Trust Region Aggressive Space Mapping Algorithm

using $\mathbf{f}^{(i)} = \mathbf{P}(\mathbf{x}_{em}^{(i)}) - \mathbf{x}_{os}^*$

solve $(\mathbf{B}^{(i)T} \mathbf{B}^{(i)} + \mathbf{I} \mathbf{I}) \mathbf{h}^{(i)} = -\mathbf{B}^{(i)T} \mathbf{f}^{(i)}$ for $\mathbf{h}^{(i)}$

this corresponds to minimizing $\|\mathbf{f}^{(i)} + \mathbf{B}^{(i)} \mathbf{h}^{(i)}\|_2^2$ subject to

$\|\mathbf{h}^{(i)}\|_2 \leq \mathbf{d}$ where \mathbf{d} is the size of the trust region

\mathbf{I} , which correlates to \mathbf{d} can be determined (*Moré et al., 1983*)

single point parameter extraction is performed at the new point

$\mathbf{x}_{em}^{(i+1)} = \mathbf{x}_{em}^{(i)} + \mathbf{h}^{(i)}$ to get $\mathbf{f}^{(i+1)}$

if $\mathbf{f}^{(i+1)}$ satisfies a certain success criterion for the reduction in the l_2 norm of the vector \mathbf{f} , the point $\mathbf{x}_{em}^{(i+1)}$ is accepted and the matrix $\mathbf{B}^{(i)}$ is updated using Broyden's update

otherwise a temporary point is generated using $\mathbf{x}_{em}^{(i+1)}$ and $\mathbf{f}^{(i+1)}$ and is added to the set of points to be used for multi-point parameter extraction

a new $\mathbf{f}^{(i+1)}$ is obtained through multi-point parameter extraction



Trust Region Aggressive Space Mapping Algorithm

the last three steps are repeated until a success criterion is satisfied or the step is declared a failure

step failure has two forms

- (1) f may approach a limiting value without satisfying the success criterion or
- (2) the number of fine model points simulated since the last successful step reaches $n+1$

Case (1): the parameter extraction is trusted but the linearization used is suspect; the size of the trust region is decreased and a new point $\mathbf{x}_{em}^{(i+1)}$ is obtained

Case (2): sufficient information is available for an approximation to the Jacobian of the fine model responses w.r.t. the fine model parameters used to predict the new point $\mathbf{x}_{em}^{(i+1)}$

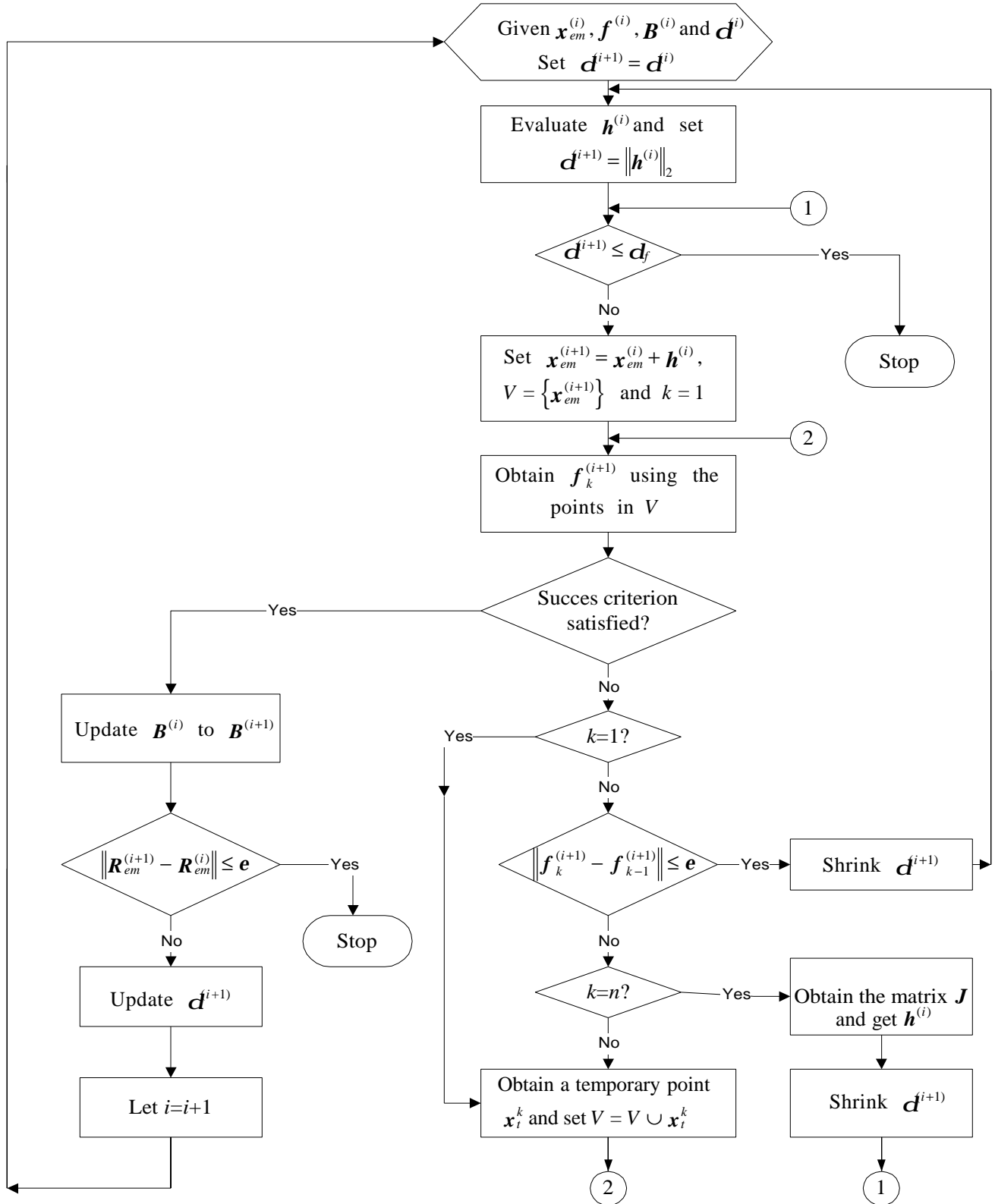
the mapping between the two spaces is exploited in the parameter extraction step by solving

$$\underset{\mathbf{x}_{os}}{\text{minimize}} \left\| \mathbf{R}_{os}(\mathbf{x}_{os} + \mathbf{B}^{(i)}(\mathbf{x} - \mathbf{x}_{em}^{(i+1)})) - \mathbf{R}_{em}(\mathbf{x}) \right\|$$

simultaneously for a set of points \mathbf{x}



Flow Chart

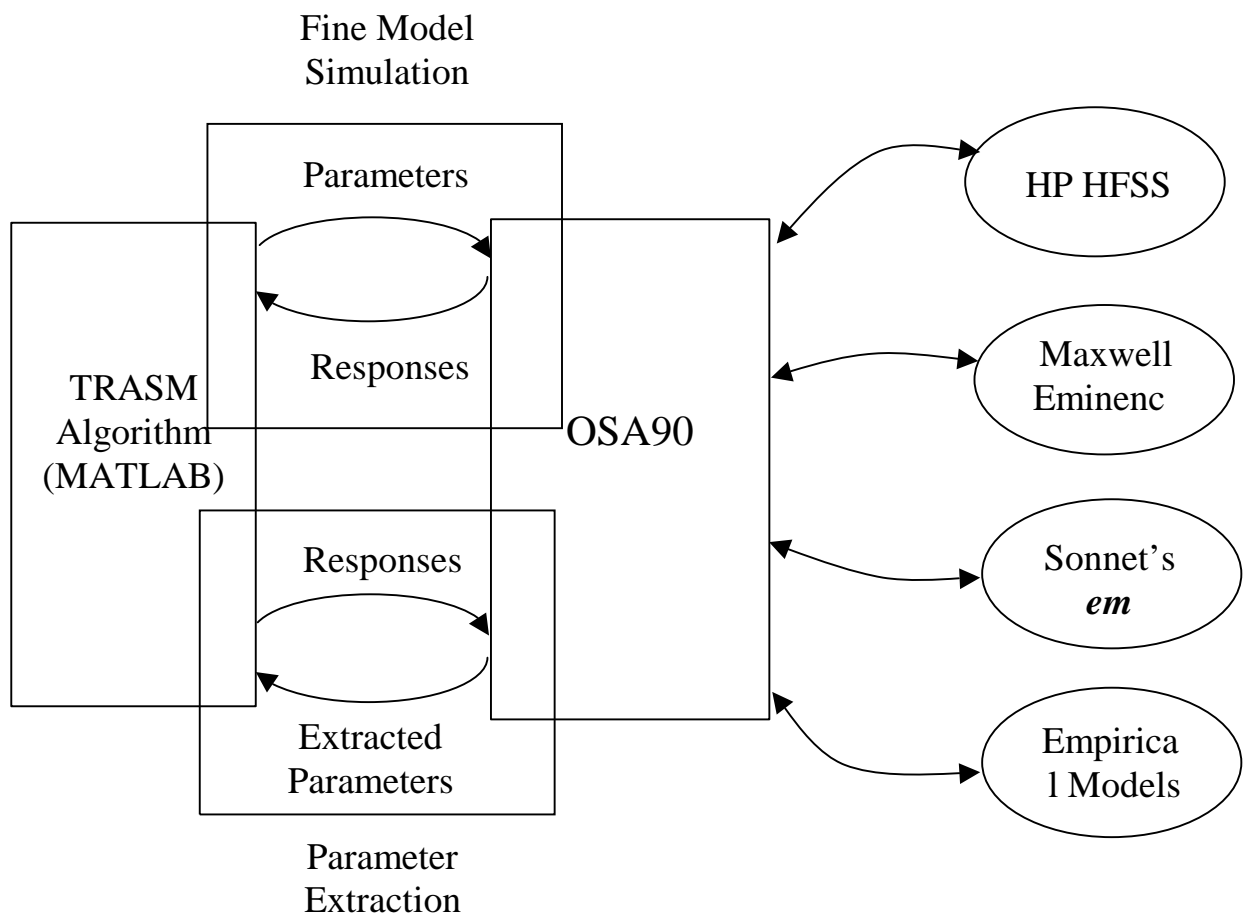




The Current Implementation

The algorithm is currently implemented in MATLAB

OSA90 is used as a platform for the multi-point parameter extraction and for the fine model simulations

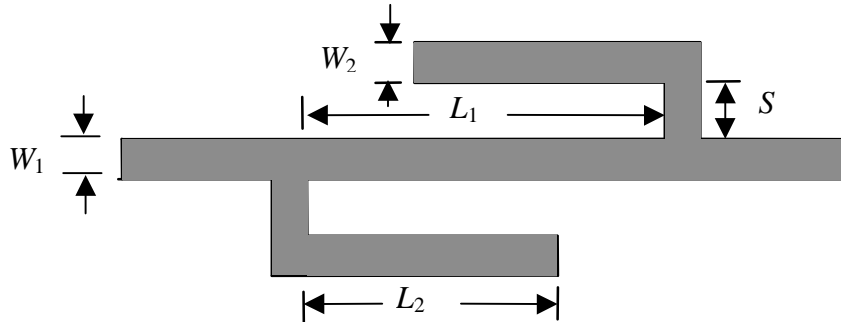


TRASM : Trust Region Aggressive Space Mapping



Double-Folded Stub Microstrip Filter

(Rautio, 1992)



passband specifications: $|S_{21}| \geq -3$ dB for $f \leq 9.5$ GHz and 16.5 GHz $\leq f$

stopband specifications: $|S_{21}| \leq -30$ dB for 12 GHz $\leq f \leq 14$ GHz

designable parameters: L_1 , L_2 and S ; $W_1=W_2=4.8$ mil

substrate thickness is 5 mil and the relative dielectric constant is assumed to be 9.9

coarse model: a coarse grid Sonnet *em* model with $\Delta x = \Delta y = 4.8$ mil

fine model: a fine grid Sonnet *em* model with $\Delta x = \Delta y = 1.6$ mil

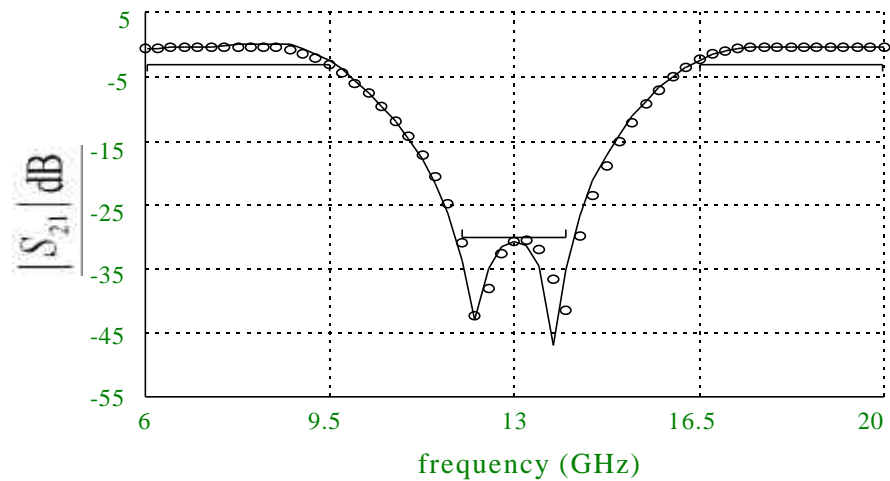
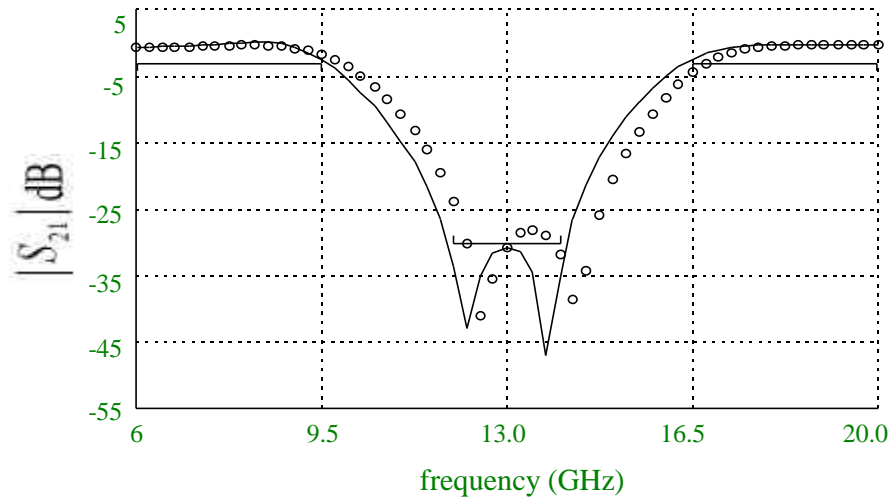
final design is obtained in 2 TRASM iterations, requiring 5 fine model points

OSA's Empipe linear interpolation is used to simulate off-grid points



Double-Folded Stub Microstrip Filter Responses

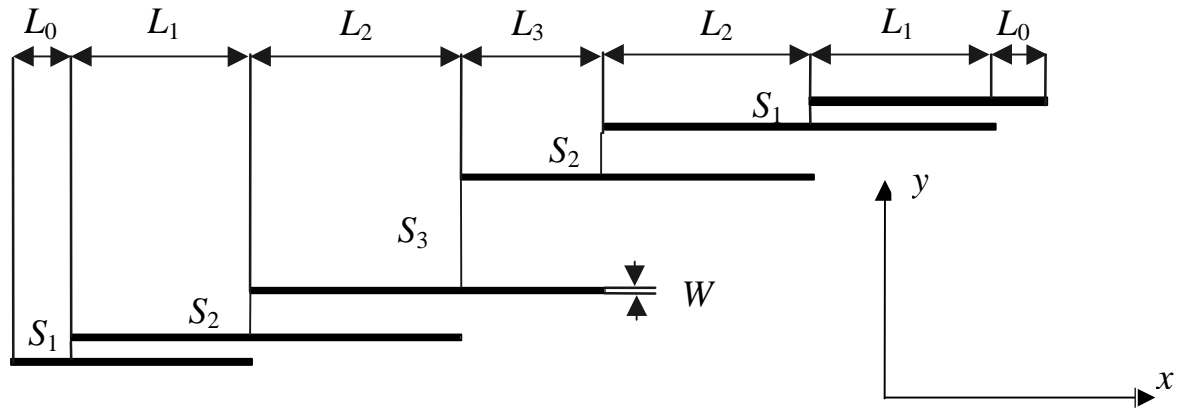
the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





High-Temperature Superconducting Filter

(Westinghouse, 1993)



20 mil thick substrate

the dielectric constant is 23.4

passband specifications: $|S_{21}| \geq 0.95$ for $f \leq 3.967$ GHz and 4.099 GHz $\leq f$

stopband specifications: $|S_{21}| \leq 0.05$ for 4.008 GHz $\leq f \leq 4.058$ GHz

designable parameters L_1 , L_2 , L_3 , S_1 , S_2 and S_3 ; L_0 and W are kept fixed

coarse model exploits the empirical models of microstrip lines, coupled lines and open stubs available in OSA90/hope



High-Temperature Superconducting Filter Fine Model

the fine model employs a fine-grid Sonnet *em* simulation

the *x* and *y* grid sizes for *em* are 1.0 and 1.75 mil

100 elapsed minutes are needed for *em* analysis at single frequency on a Sun SPARCstation 10

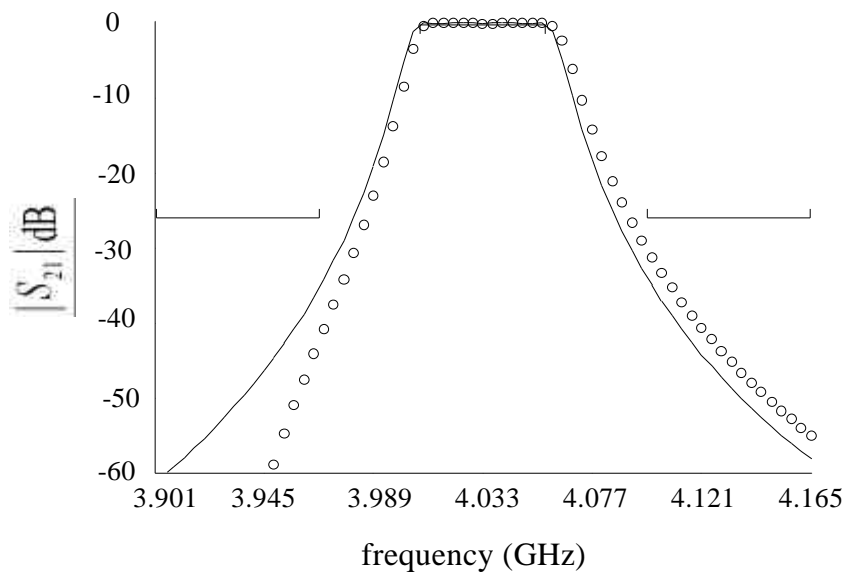
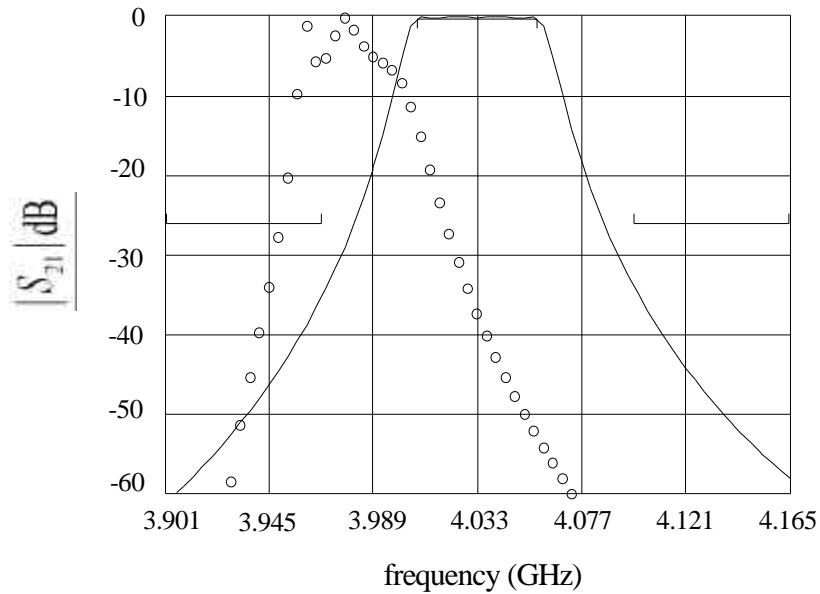
final design is obtained in 5 TRASM iterations, requiring 8 *em* simulations

15 frequency points are used per *em* simulation



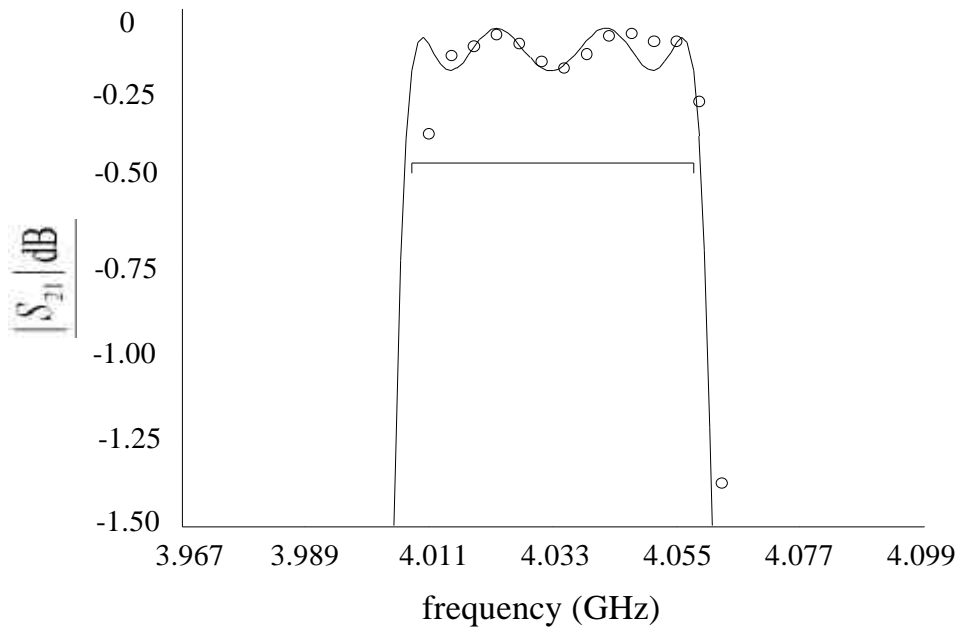
High-Temperature Superconducting Filter Responses

the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





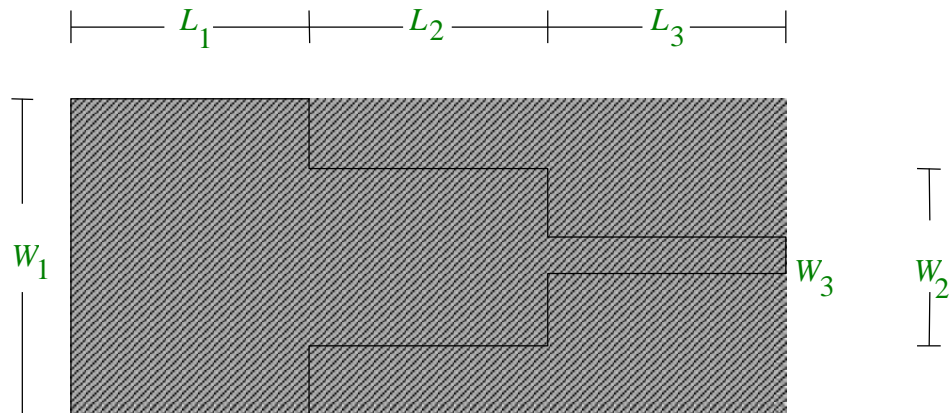
Passband Details for the High-Temperature Superconducting Filter





Three-Section 3:1 Microstrip Transformer

(Bandler et al., 1994)



source and load impedances 50Ω and 150Ω

design constraints $|S_{11}| \leq 0.11$ for the frequency range
 $5 \text{ GHz} \leq f \leq 15 \text{ GHz}$.

designable parameters: W_1 , W_2 , W_3 , L_1 , L_2 and L_3

coarse model: ideal transmission-line model supplied by
OSA90/hope

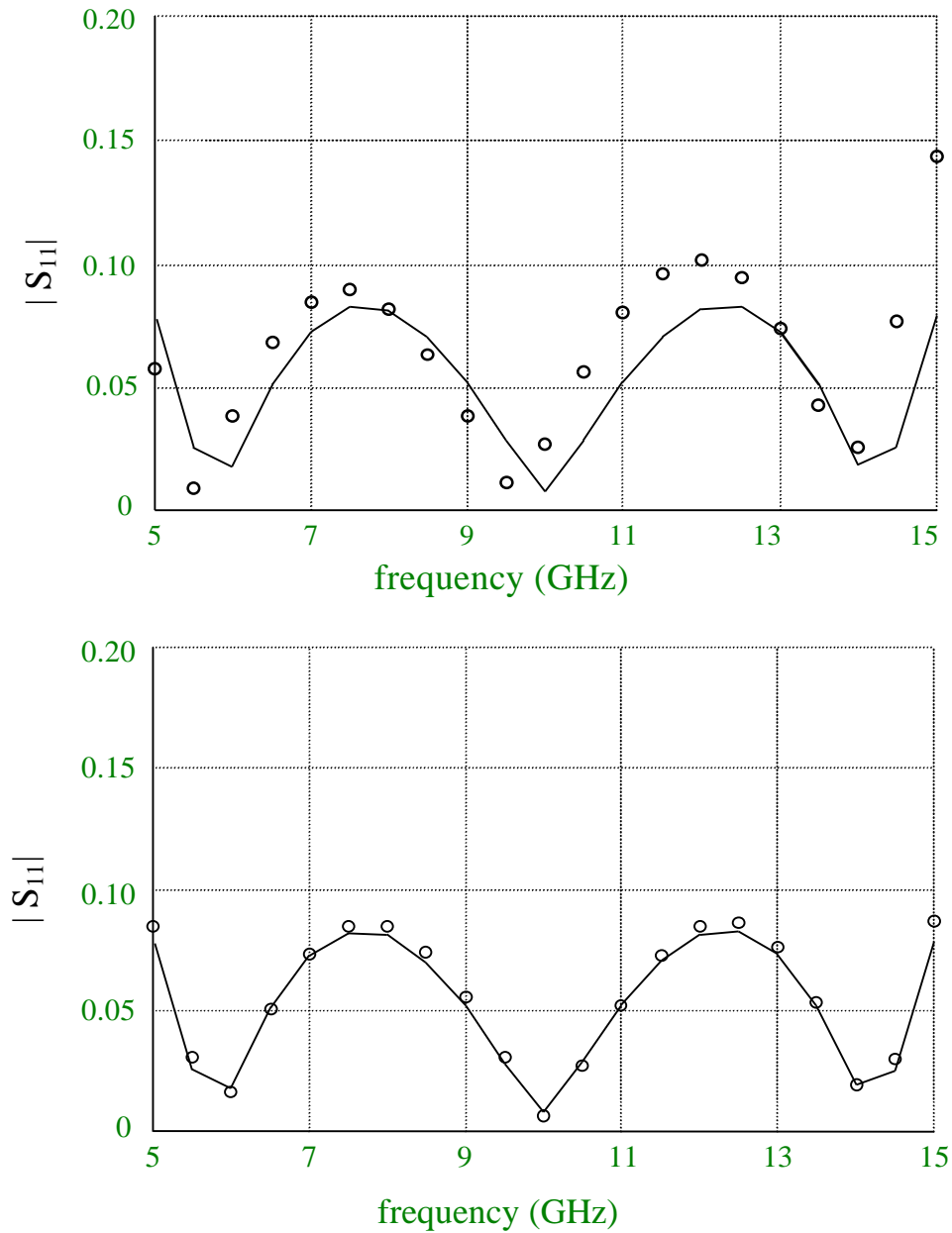
fine model exploits microstrip and microstrip step models
supplied by OSA90/hope

the final solution is obtained in 2 TRASM iterations, requiring 6
fine model simulations



Three-Section 3:1 Microstrip Transformer Responses

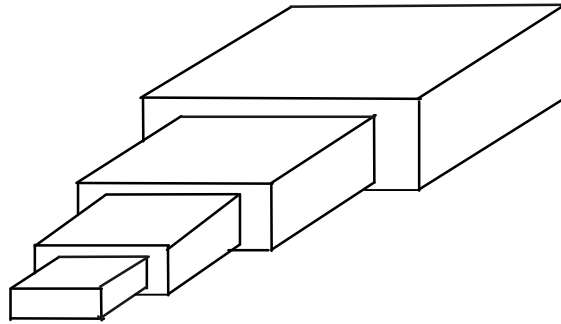
the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





Two-Section Waveguide Transformer

(Bandler et al., 1996)



design specifications: $VSWR \leq 1.01$ for $5.8 \text{ GHz} \leq f \leq 6.6 \text{ GHz}$

optimizable parameters: length and height of each waveguide section

coarse model: an analytical model that neglects junction discontinuity

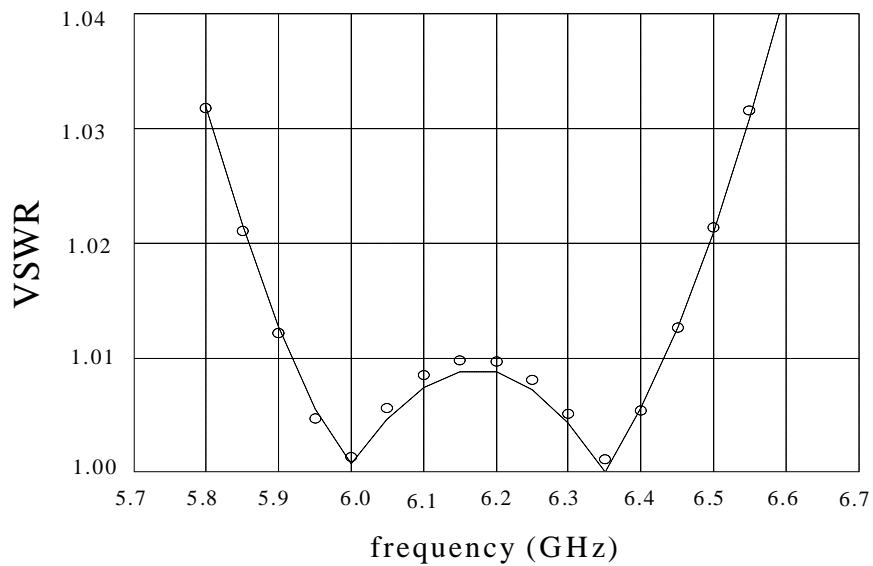
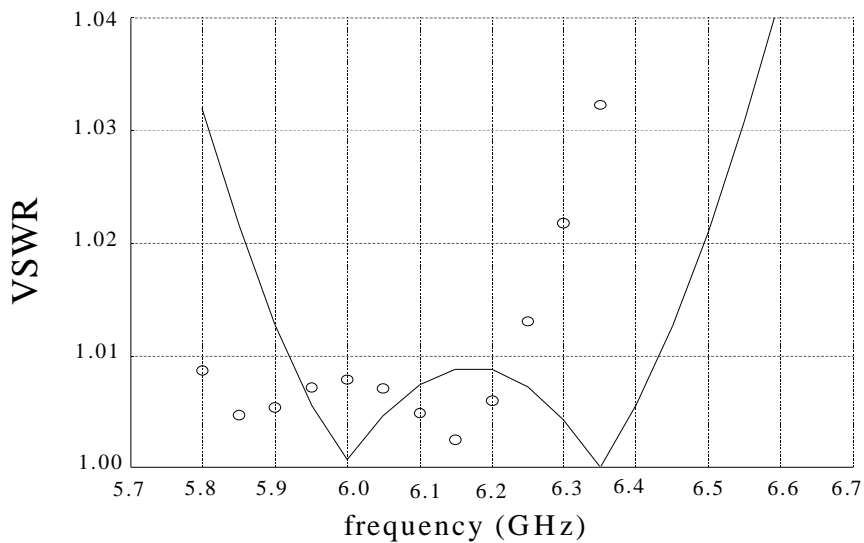
fine model exploits the 3D solver Maxwell Eminence

the optimal design is obtained in 3 TRASM iterations requiring 5 Maxwell Eminence simulations



Two-Section Waveguide Transformer Responses

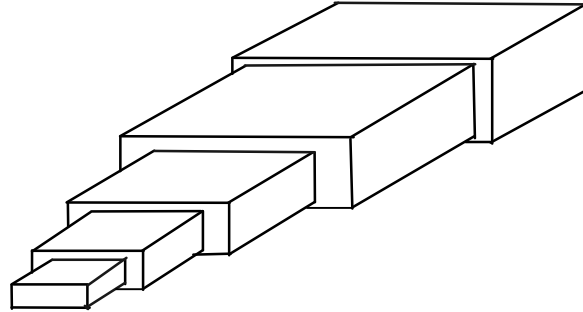
the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





Three-Section Waveguide Transformer

(Bandler et al., 1996)



design specifications: $VSWR \leq 1.04$ for $5.7 \text{ GHz} \leq f \leq 7.2 \text{ GHz}$

optimizable parameters: length and height of each waveguide section

coarse model: an analytical model that neglects junction discontinuity

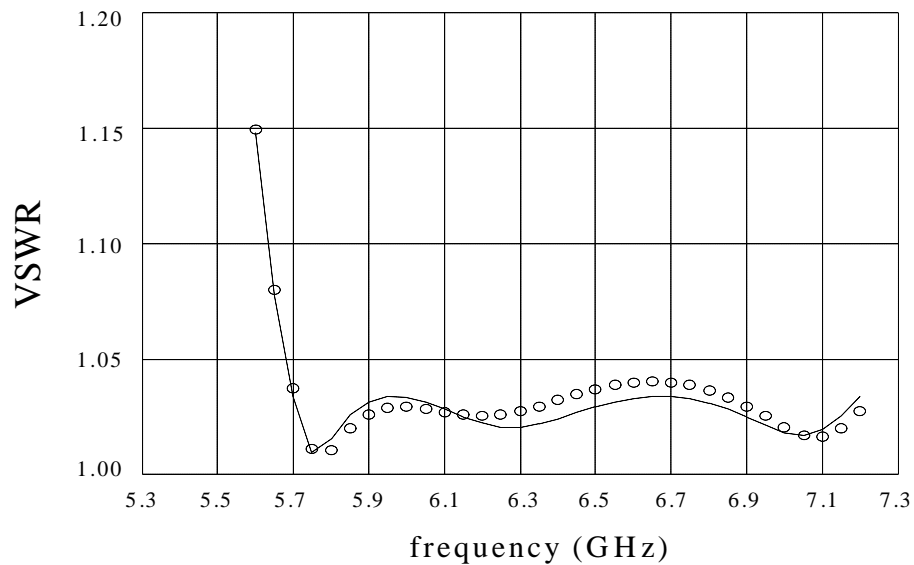
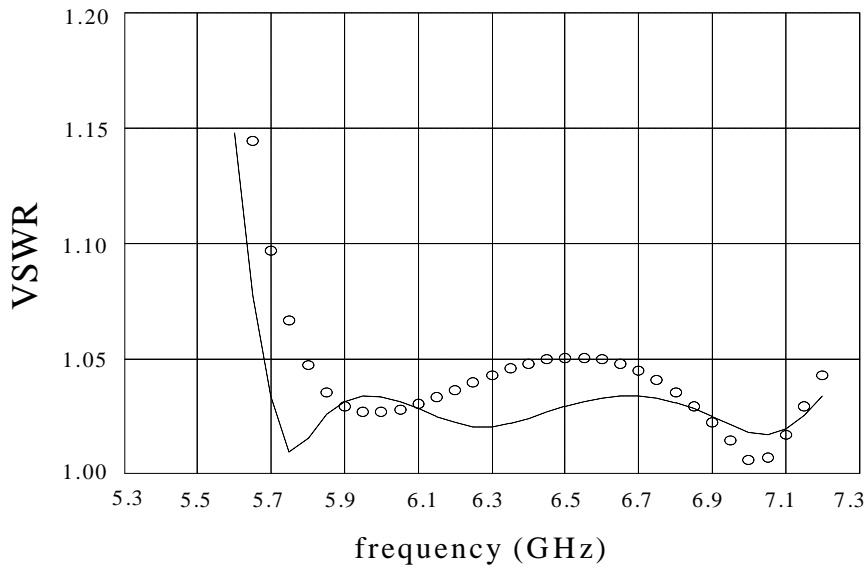
fine model exploits the 3D solver Maxwell Eminence

the optimal design is obtained in 2 TRASM iterations requiring 9 Maxwell Eminence simulations



Three-Section Waveguide Transformer Responses

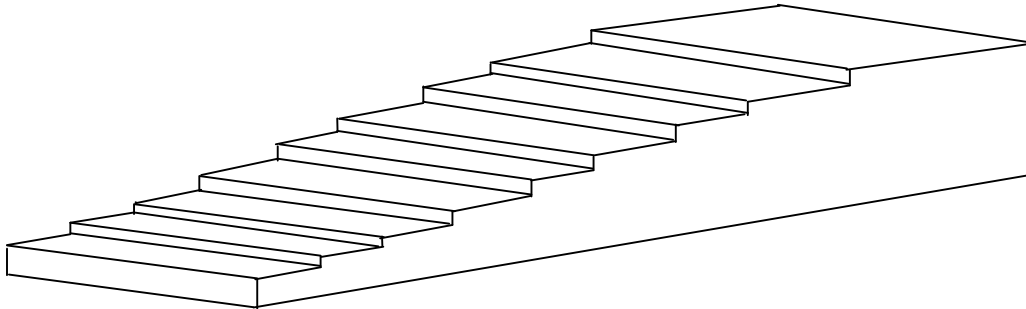
the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





Seven-Section Waveguide Transformer

(Bandler et al., 1996)



design specifications: $VSWR \leq 1.01$ for $1.06 \text{ GHz} \leq f \leq 1.8 \text{ GHz}$

optimizable parameters: length and height of each waveguide section

coarse model: an analytical model that neglects junction discontinuities

fine model : an analytical model that considers junction discontinuities

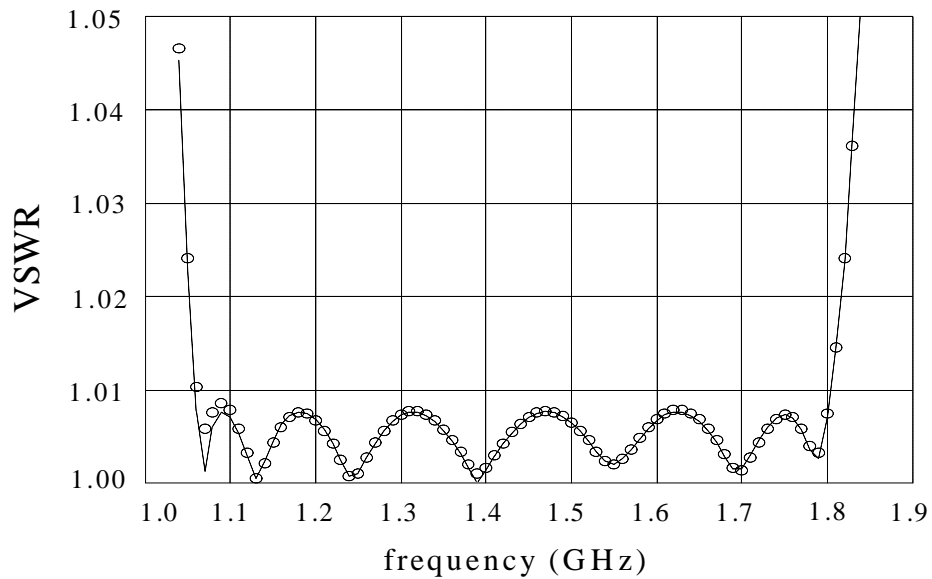
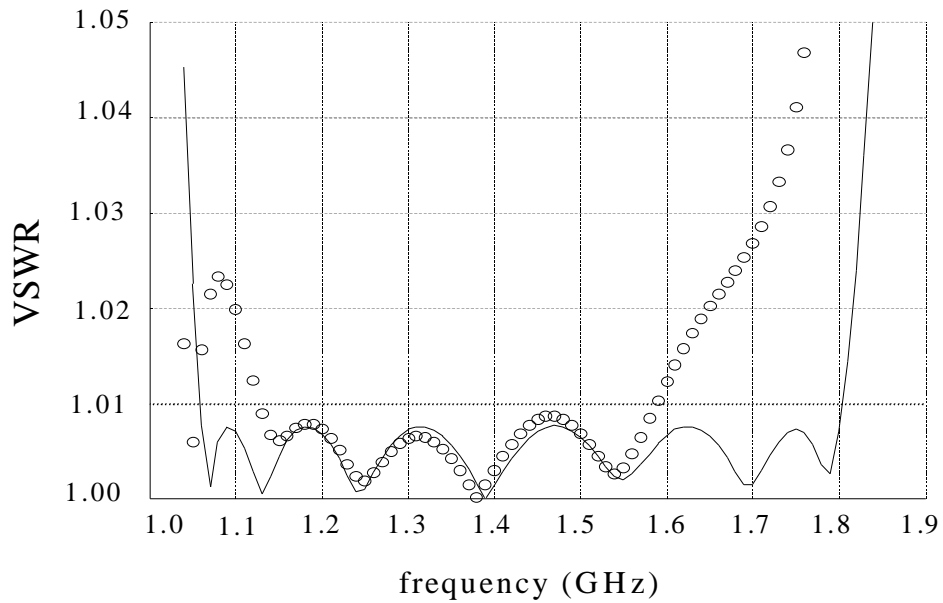
the optimal design is obtained in 3 TRASM iterations requiring 6 fine model simulations

21 frequency points were used per simulation



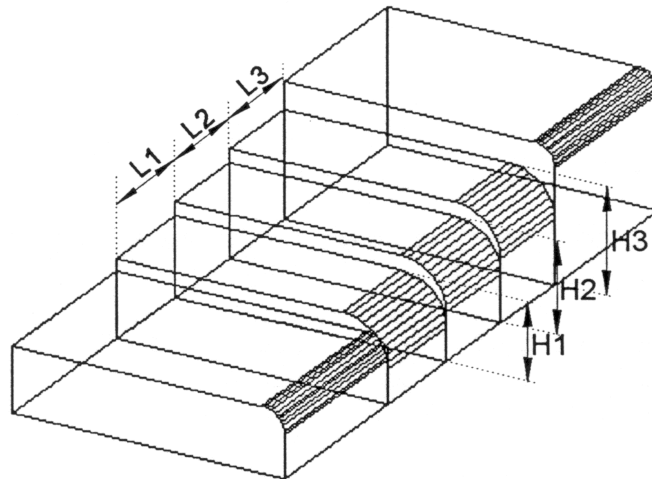
Seven-Section Waveguide Transformer Responses

the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





Three-Section Waveguide Transformer, Rounded Corners (*Empipe3D manual, 1997*)



design specifications: $|S_{11}| \leq -30\text{dB}$ for $9.5\text{ GHz} \leq f \leq 15\text{ GHz}$

impedance matching between WR-75 half height and WR-75 full height waveguides

designable variables: the height and length of each section

fine (EM) model: HP HFSS Version 5.0

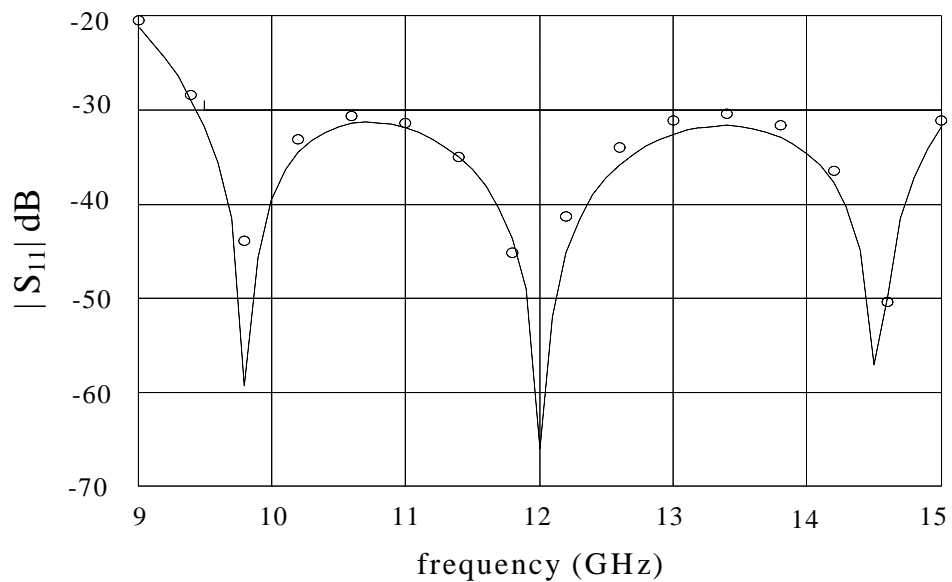
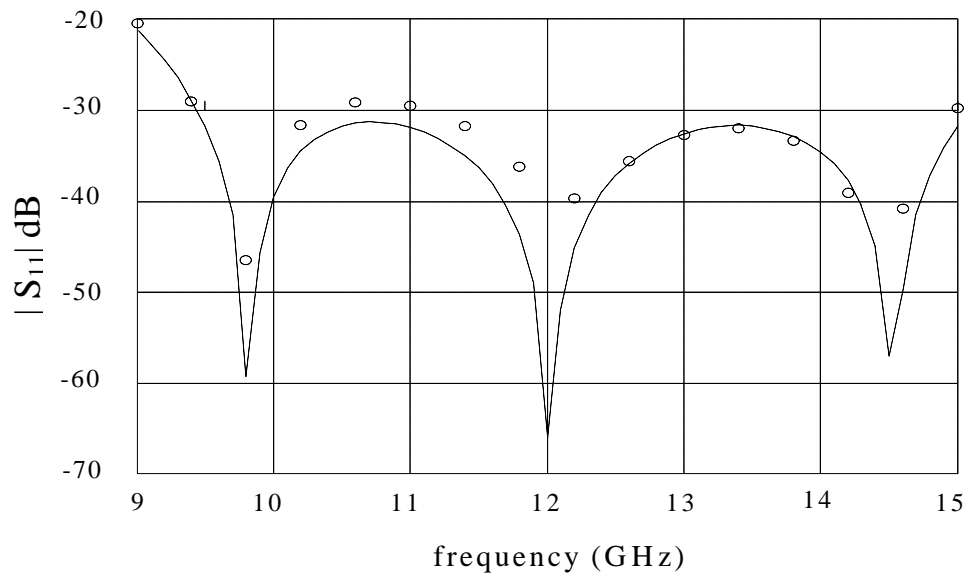
coarse model: ideal analytical model (*Bandler, 1969*)

the optimal design was obtained in 1 TRASM iteration,
requiring 7 fine model simulations by HP HFSS



Three-Section Waveguide Transformer, Rounded Corners

the optimal coarse model (—) response and the fine model response (o) at the initial and final designs





Conclusions

we present a novel Trust Region Aggressive Space Mapping algorithm for the optimization of microwave circuits

TRASM integrates a trust region methodology with the Aggressive Space Mapping technique

a recursive multi-point parameter extraction step is exploited to improve the uniqueness of the parameter extraction step

the multi-point parameter extraction exploits the available information about the mapping between the two spaces to improve the uniqueness of the step

a number of examples were successfully optimized using the proposed algorithm

the algorithm is currently implemented in MATLAB