Expanded Space Mapping Optimization of Microwave Circuits Exploiting Preassigned Parameters

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outline

Space Mapping concept

Key Preassigned Parameters (KPP)

coarse model decomposition

Expanded Space Mapping Design Framework (ESMDF) algorithm

examples





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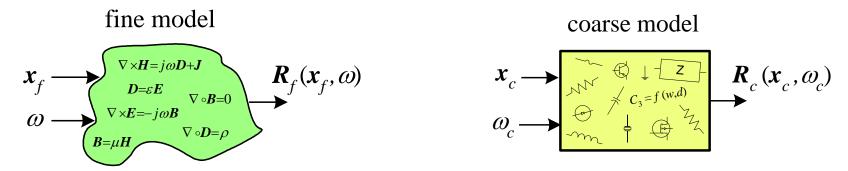
examples

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Space Mapping Concept

(*Bandler et. al., 1994*)



find

$$\begin{bmatrix} \boldsymbol{x}_c \\ \boldsymbol{\omega}_c \end{bmatrix} = \boldsymbol{P}(\boldsymbol{x}_f, \boldsymbol{\omega})$$

such that

$$\mathbf{R}_c(\mathbf{x}_c, \omega_c) \approx \mathbf{R}_f(\mathbf{x}_f, \omega)$$



the KPP are assumed to be non-optimizable

examples: dielectric constant, substrate height, etc.

the coarse model is very sensitive to KPP

the coarse model is calibrated to match the fine model by tuning the KPP

our algorithm establishes a mapping from some optimizable parameters to KPP



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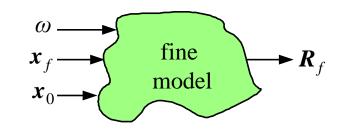
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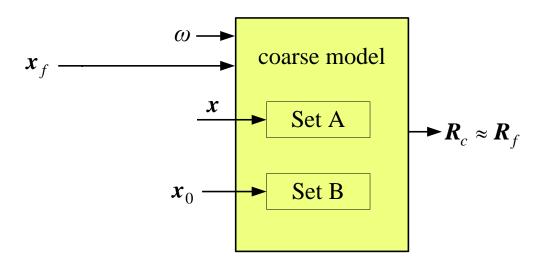
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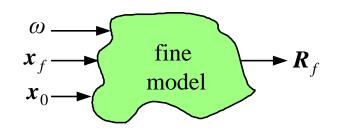


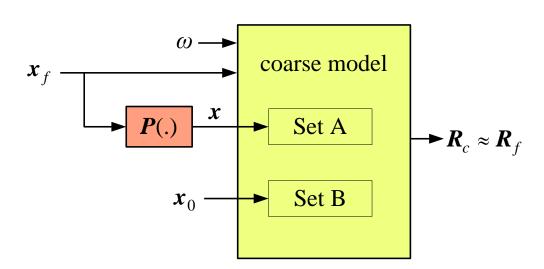


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Key Preassigned Parameters (KPP)



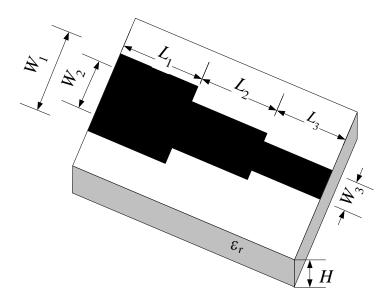


$$x = P(x_r)$$

$$\boldsymbol{x}_f = [\boldsymbol{x}_r^T \quad \boldsymbol{x}_s^T]^T$$

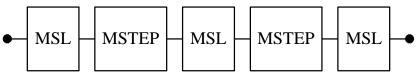
$$x = c + B_r x_r$$





$$\varepsilon_r = 9.7$$
, $H = 25$ mil

comp. #1 comp. #2 comp. #3 comp. #4 comp. #5



$$\boldsymbol{x}_f = [W_1 \quad W_2 \quad W_3 \quad L_1 \quad L_2 \quad L_3]^T$$

$$\boldsymbol{x}_r = [W_1 \quad W_2 \quad W_3]^T$$

$$\boldsymbol{x} = [\boldsymbol{x}_1^T \ \boldsymbol{x}_3^T \ \boldsymbol{x}_5^T]^T \quad \boldsymbol{x}_i = [\boldsymbol{\varepsilon}_{ri} \ H_i]^T$$

$$x = c + B_r x_r$$





 x_i represents the KPP of the *i*th component, $i \in I = \{1, 2, ..., N\}$

N is the number of coarse model components

Set A: contains "relevant" coarse model components

Set B: contains coarse model components for which the coarse model

is insensitive to their KPP





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for all $i \in I = \{1, 2, ..., N\}$ evaluate Step 1

$$S_i = \left\| \left(\frac{\partial \mathbf{R}_c^{\mathrm{T}}}{\partial \mathbf{x}_i} \mathbf{D} \right)^{\mathrm{T}} \right\|_F, \quad \mathbf{D} = \operatorname{diag}(\mathbf{x}_0)$$

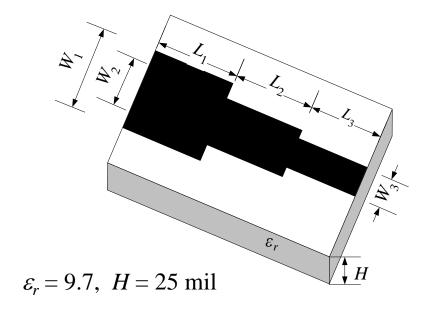
Step 2 evaluate

$$\hat{S}_i = \frac{S_i}{\max_{j \in I} \{S_j\}}, i \in I$$

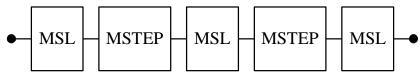
put the *i*th component in Set A if $\hat{S}_i \ge \beta$ Step 3 otherwise put it in Set B ($\beta = 0.2$)



example: 3:1 microstrip transformer



comp. #1 comp. #2 comp. #3 comp. #4 comp. #5



$$\mathbf{x}_i = [\varepsilon_{ri} \quad H_i]^T, \quad i = 1, \dots, 5$$

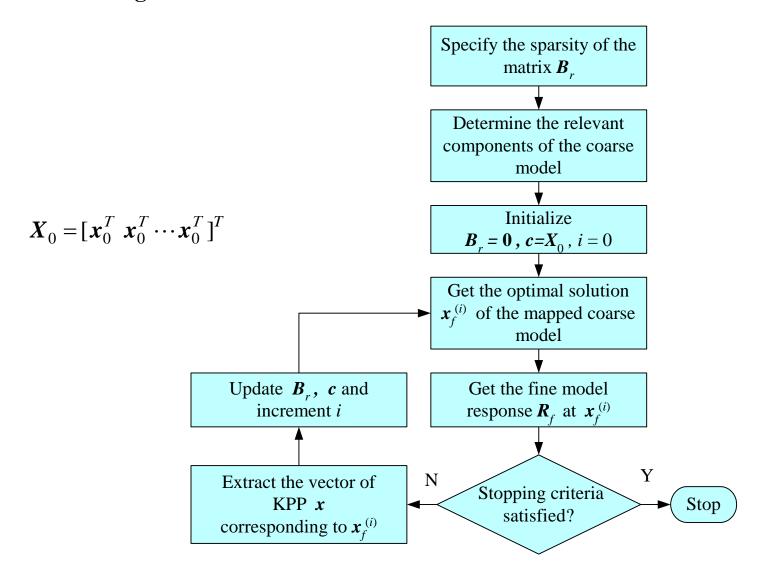
$$S_i = \left\| \left(\frac{\partial \boldsymbol{R}_c^{\mathrm{T}}}{\partial \boldsymbol{x}_i} \boldsymbol{D} \right)^{\mathrm{T}} \right\|_F$$

Component #	\hat{S}_i
1	1
2	0.05
3	0.39
4	0.04
5	0.77

hence
$$\mathbf{x} = [\mathbf{x}_1^T \ \mathbf{x}_3^T \ \mathbf{x}_5^T]^T$$



ESMDF Algorithm





Expanded Space Mapping Optimization Algorithm

mapped coarse model optimization

$$\boldsymbol{x}_f^{(i)} = \arg\min_{\boldsymbol{x}_f} U(\boldsymbol{R}_c(\boldsymbol{x}_f, \boldsymbol{x}))$$

$$\boldsymbol{x} = \boldsymbol{B}_r \ \boldsymbol{x}_r + \boldsymbol{c}$$

$$\boldsymbol{x}_f = [\boldsymbol{x}_r^T \quad \boldsymbol{x}_s^T]^T$$



Expanded Space Mapping Optimization Algorithm

mapped coarse model optimization exploiting trust region methodology

$$\boldsymbol{h} = \arg\min_{\boldsymbol{h}} U(\boldsymbol{R}_c(\boldsymbol{x}_f^{(i-1)} + \boldsymbol{h}, \boldsymbol{B}_r \ \boldsymbol{x}_r^{(i-1)} + \boldsymbol{c}))$$
subject to $\|\Lambda \boldsymbol{h}\| \le \delta$

successful iteration

$$\boldsymbol{x}_{f}^{(i)} = \begin{cases} \boldsymbol{x}_{f}^{(i-1)} + \boldsymbol{h} & \text{if } U(\boldsymbol{R}_{f}(\boldsymbol{x}_{f}^{(i-1)} + \boldsymbol{h})) < U(\boldsymbol{R}_{f}(\boldsymbol{x}_{f}^{(i-1)})) \\ \boldsymbol{x}_{f}^{(i-1)} & \text{otherwise} \end{cases}$$



Expanded Space Mapping Optimization Algorithm

KPP extraction

$$\boldsymbol{x}^{(i)} = \arg\min_{\boldsymbol{x}} \left\| \boldsymbol{R}_f(\boldsymbol{x}_f^{(i)}) - \boldsymbol{R}_c(\boldsymbol{x}_f^{(i)}, \boldsymbol{x}) \right\|$$

stopping criteria

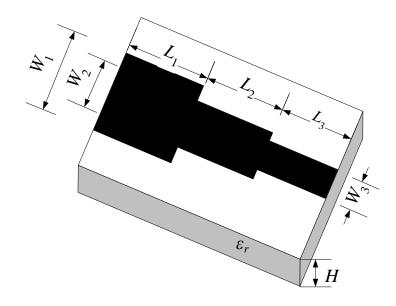
$$\left\| \boldsymbol{x}_{f}^{(i)} - \boldsymbol{x}_{f}^{(i-1)} \right\| \leq \varepsilon_{1}$$

$$\| \boldsymbol{R}_f(\boldsymbol{x}_f^{(i)}) - \boldsymbol{R}_c(\boldsymbol{x}_f^{(i)}, \boldsymbol{x}^{(i-1)} + \boldsymbol{B}_r^{(i-1)} \boldsymbol{h}_r^{(i)}) \| \le \varepsilon_2$$

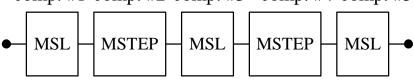
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3:1 Microstrip Transformer



comp. #1 comp. #2 comp. #3 comp. #4 comp. #5



load impedance is 50Ω

source impedance is 150 Ω

"fine" model: Sonnet's *em*parameterized by OSA's Empipe

"coarse" model: OSA90/hope

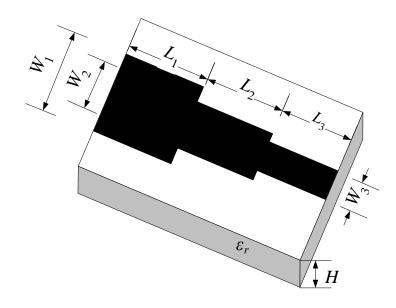
specifications

 $|S_{11}| \le -20 \text{ dB for 5 GHz} \le \omega \le 15 \text{ GHz}$

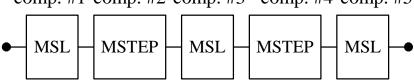
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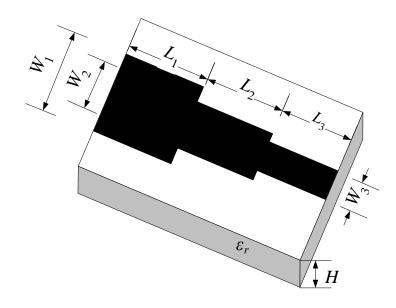
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comp. #1 comp. #2 comp. #3 comp. #4 comp. #5

MSL MSTEP MSL MSTEP MSL

$$\boldsymbol{x}_f = \begin{bmatrix} W_1 & W_2 & W_3 & L_1 & L_2 & L_3 \end{bmatrix}^T$$

$$\boldsymbol{x}_r = [W_1 \quad W_2 \quad W_3]^T$$

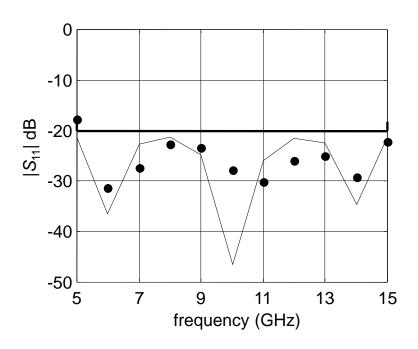
$$\boldsymbol{x} = [\boldsymbol{x}_1^T \ \boldsymbol{x}_3^T \ \boldsymbol{x}_5^T]^T \quad \boldsymbol{x}_i = [\boldsymbol{\varepsilon}_{ri} \ H_i]^T$$

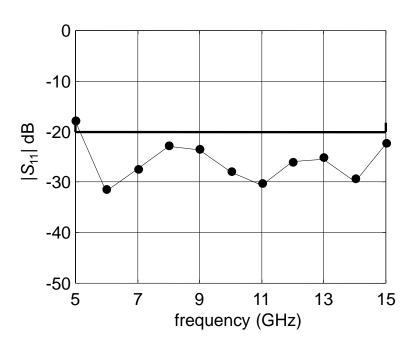
$$x = c + B_r x_r$$

$$\begin{bmatrix} \varepsilon_{r1} \\ H_1 \\ \varepsilon_{r3} \\ H_3 \\ \varepsilon_{r5} \\ H_5 \end{bmatrix} = \begin{array}{cccc} c & + & \begin{bmatrix} x & 0 & 0 \\ x & 0 & 0 \\ 0 & x & 0 \\ 0 & x & 0 \\ 0 & 0 & x \\ 0 & 0 & x \end{bmatrix} \begin{bmatrix} W_1 \\ W_2 \\ W_3 \end{bmatrix}$$



initial iteration



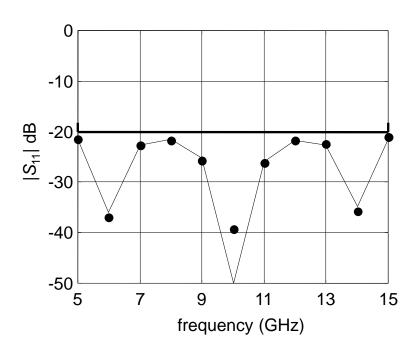


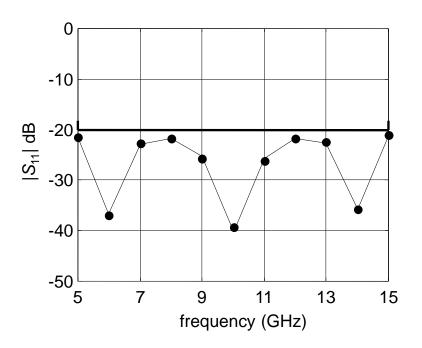
before KPP extraction

after KPP extraction



next iteration



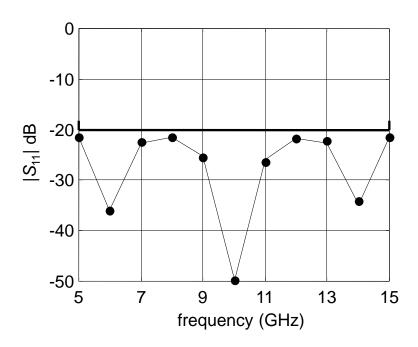


before KPP extraction

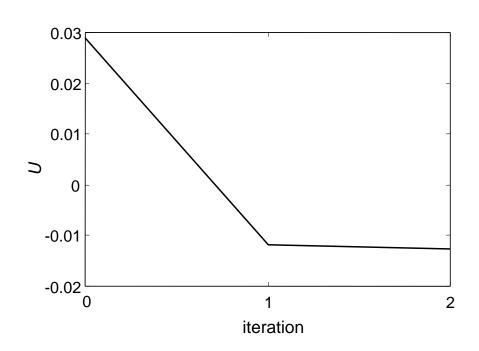
after KPP extraction



final iteration



fine model objective function

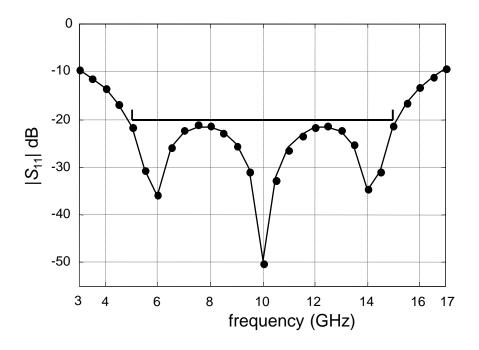


elapsed time by the ESMDF algorithm: 17 min



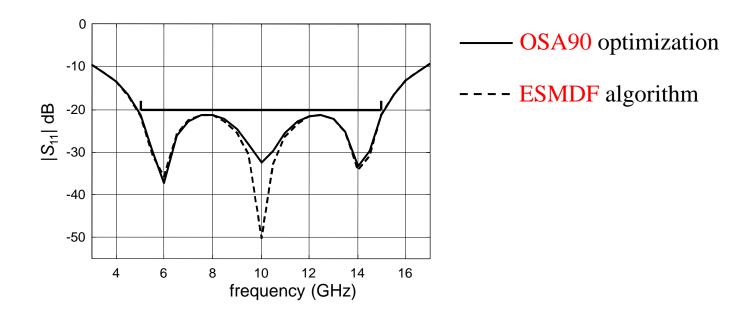


detailed frequency sweep of the optimal response





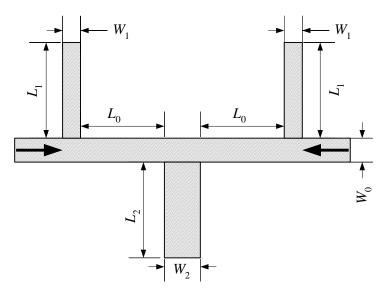
3:1 Microstrip Transformer Direct EM Optimization



elapsed time by OSA90 minimax optimization (using quadratic interpolation): 153 min elapsed time by the ESMDF algorithm: 17 min



Microstrip Bandstop Filter with Open Stubs



"fine" model: Momentum (Agilent EEsof EDA)



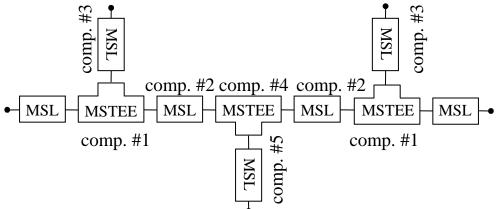
"coarse" model: OSA90/hope



specifications

 $|S_{21}| \ge -1$ dB for $\omega \ge 12$ GHz and $\omega \le 8$ GHz

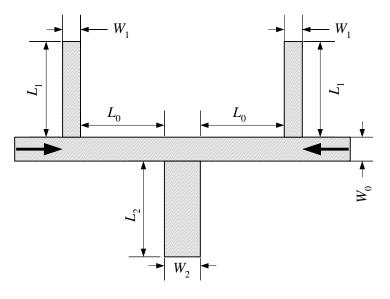
 $|S_{21}| \le -25 \text{ dB for } 9 \text{ GHz} \le \omega \le 11 \text{ GHz}$



Simulation Optimization Systems Research Laboratory



Microstrip Bandstop Filter with Open Stubs



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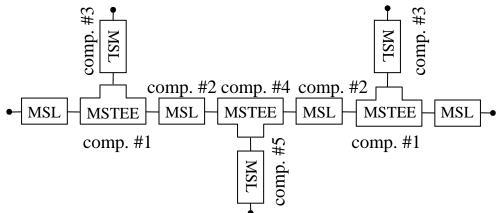
"coarse" model: OSA90/hope



specifications

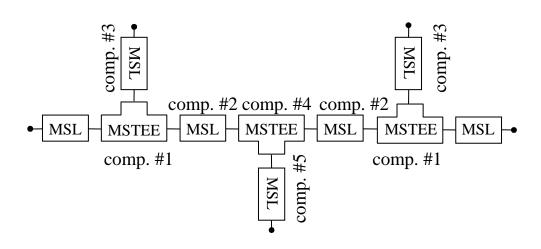
 $|S_{21}| \ge -1$ dB for $\omega \ge 12$ GHz and $\omega \le 8$ GHz

$$|S_{21}| \le -25 \text{ dB for } 9 \text{ GHz} \le \omega \le 11 \text{ GHz}$$





coarse model decomposition



$$\varepsilon_r = 9.4$$
, $H = 25$ mil

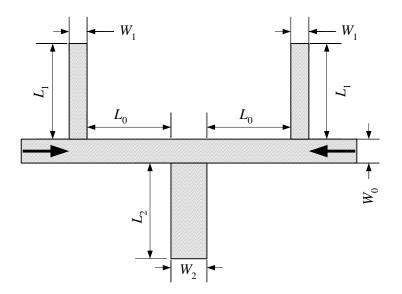
$$\mathbf{x}_i = \begin{bmatrix} \varepsilon_{ri} & H_i \end{bmatrix}^T, \quad i = 1, \dots, 5$$

$$S_i = \left\| \left(\frac{\partial \boldsymbol{R}_c^{\mathrm{T}}}{\partial \boldsymbol{x}_i} \boldsymbol{D} \right)^{\mathrm{T}} \right\|_{\boldsymbol{P}}$$

Component #	\hat{S}_i	
1	0.1420	
2	0.6359	
3	0.8395	
4	0.1858	
5	1.0000	

hence
$$\mathbf{x} = [\mathbf{x}_2^T \ \mathbf{x}_3^T \ \mathbf{x}_5^T]^T$$





$$\boldsymbol{x}_f = [W_1 \quad W_2 \quad L_0 \quad L_1 \quad L_2]^T$$

$$\boldsymbol{x}_r = [W_1 \quad W_2]^T$$

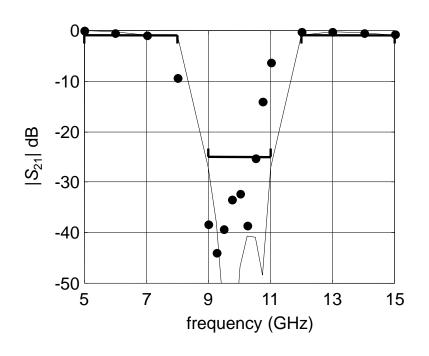
$$\mathbf{x} = [\mathbf{x}_2^T \ \mathbf{x}_3^T \ \mathbf{x}_5^T]^T \ \mathbf{x}_i = [\varepsilon_{ri} \ H_i]^T$$

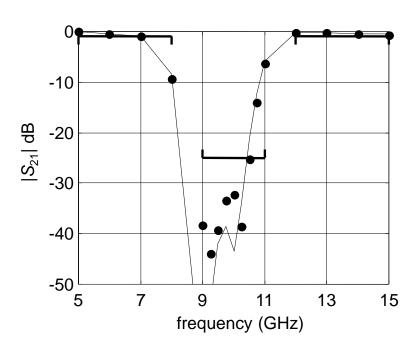
$$x = c + B_r x_r$$

$$\begin{bmatrix} \varepsilon_{r2} \\ H_2 \\ \varepsilon_{r3} \\ H_3 \\ \varepsilon_{r5} \\ H_5 \end{bmatrix} = \mathbf{c} + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ \mathbf{x} & 0 \\ \mathbf{x} & 0 \\ 0 & \mathbf{x} \\ 0 & \mathbf{x} \end{bmatrix} \begin{bmatrix} W_1 \\ W_2 \end{bmatrix}$$



initial response



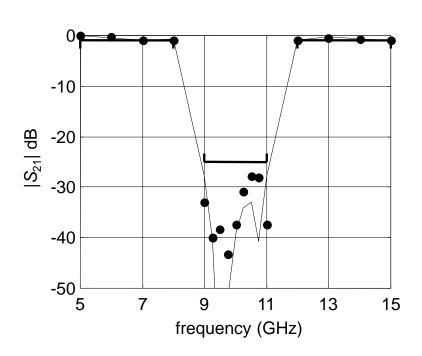


before KPP extraction

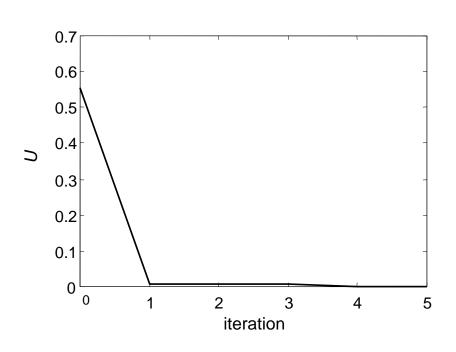
after KPP extraction



final response



fine model objective function

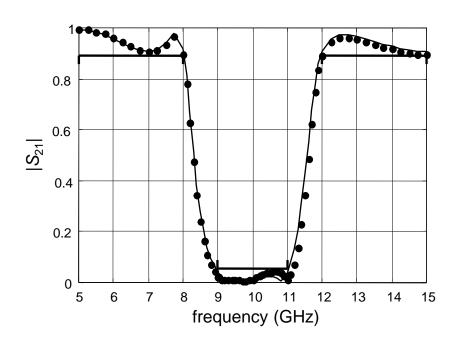


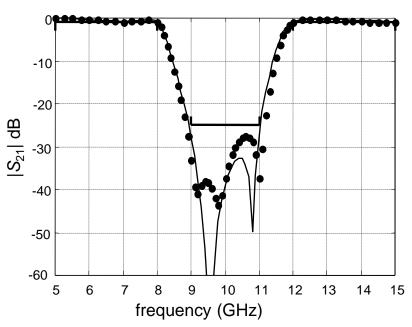
elapsed time by the **ESMDF** algorithm: 1.5 hr





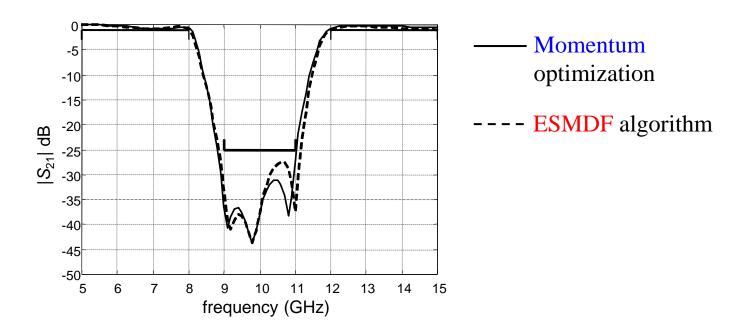
detailed frequency sweep at the optimal solution







direct optimization



elapsed time by Momentum optimization (using quadratic interpolation): 10 hr elapsed time by the ESMDF algorithm: 1.5 hr



we expand the original space mapping approach

we exploit key preassigned parameters (KPP)

we tune the KPP in "relevant components" of the coarse model to align it with the fine model

a mapping is established from the optimization variables to the KPP



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3:1 Microstrip Transformer Yield Analysis

utilize the mapped coarse model obtained at the final iteration

assume a uniform distribution with 0.25 mil tolerance on all six geometrical parameters

estimate the yield at the solution obtained by the ESMDF algorithm

mapped coarse model: 78 %

fine model: 79%



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