Space Mapping for Modeling Microwave Circuits

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Outline

Space Mapping concept

Generalized Space Mapping (GSM) tableau approach to modeling

Neural Space Mapping modeling

conclusions





The Space Mapping Concept

(Bandler et al., 1994-)







The Space Mapping Concept (continued)







Frequency Space Mapping Concept

(Bandler et. al., 1995)







find

$$\begin{bmatrix} \boldsymbol{x}_c \\ \boldsymbol{\omega}_c \end{bmatrix} = \boldsymbol{P}(\boldsymbol{x}_f, \boldsymbol{\omega})$$

such that

 $\boldsymbol{R}_{c}(\boldsymbol{x}_{c},\boldsymbol{\omega}_{c}) \approx \boldsymbol{R}_{f}(\boldsymbol{x}_{f},\boldsymbol{\omega})$





Mathematical Formulation for GSM

the *k*th mapping is given by

$$(\boldsymbol{x}_{ck}, \omega_{ck}) = \boldsymbol{P}_k(\boldsymbol{x}_f, \omega)$$

in matrix form, assuming a linear mapping

$$\begin{bmatrix} \boldsymbol{x}_{ck} \\ \boldsymbol{\omega}_{ck} \end{bmatrix} = \begin{bmatrix} \boldsymbol{c}_k \\ \boldsymbol{\delta}_k \end{bmatrix} + \begin{bmatrix} \boldsymbol{B}_k & \boldsymbol{s}_k \\ \boldsymbol{t}_k^T & \boldsymbol{\sigma}_k \end{bmatrix} \begin{bmatrix} \boldsymbol{x}_f \\ \boldsymbol{\omega} \end{bmatrix}$$

the mapping parameters { c_k , B_k , s_k , t_k , σ_k , δ_k } can be evaluated by solving the optimization problem

$$\min_{\boldsymbol{c}_k, \boldsymbol{B}_k, \boldsymbol{s}_k, \boldsymbol{t}_k, \boldsymbol{\sigma}_k, \boldsymbol{\delta}_k} \| [\boldsymbol{e}_{k1}^T \quad \boldsymbol{e}_{k2}^T \quad \cdots \quad \boldsymbol{e}_{km}^T]^T \|$$

where *m* is the number of base points selected in the fine model space and e_{kj} is an error vector given by

$$e_{kj} = R_f(x_f^{(j)}, \omega) - R_c(x_{ck}^{(j)}, \omega_{ck}), \quad j = 1, 2, ..., m$$





Starting Point and Learning Samples

we chose a unit mapping ($\mathbf{x}_c \approx \mathbf{x}_f$ and $\omega_c \approx \omega$) as the starting point for the optimization problem

2n+1 points are used for a microwave circuit with *n* design parameters







Multiple Space Mapping (MSM) Concept

MSM for Device Responses (MSMDR)







Multiple Space Mapping (MSM) Concept

MSM for Frequency Intervals (MSMFI)







the fine and coarse models







the region of interest

 $15 \text{ mil} \le H \le 25 \text{ mil}$ $2 \text{ mil} \le X \le 10 \text{ mil}$ $15 \text{ mil} \le Y \le 25 \text{ mil}$ $8 \le \varepsilon_r \le 10$

the frequency range is 2 GHz to 20 GHz with a step of 2 GHz

the number of base points is 9, the number of test points is 50

the widths W of the input lines track H so that their characteristic impedance is 50 ohm

 $W_1 = W/3$

 W_2 is suitably constrained





MSMFI is developed to enhance the accuracy of the coarse model

our algorithm determined two intervals: 2-16 GHz and 16-20 GHz

	2 GHz to 16 GHz	16 GHz to 20 GHz
В	$\begin{bmatrix} 1.04 & 0.07 & 0.01 & 0.08 & -0.06 & 0.00 & 0.22 \\ 0.00 & 0.89 & 0.00 & -0.07 & -0.20 & 0.06 & -0.03 \\ -0.00 & 0.07 & 0.99 & 0.04 & -0.12 & 0.01 & -0.06 \\ -0.04 & 0.00 & -0.01 & 0.97 & 0.10 & -0.06 & -0.27 \\ 0.01 & 0.04 & 0.00 & 0.03 & 0.99 & -0.05 & -0.03 \\ -0.13 & -0.05 & -0.04 & -0.16 & 0.12 & 0.99 & 0.62 \\ -0.08 & 0.12 & -0.03 & 0.00 & -0.07 & 0.03 & 0.83 \end{bmatrix}$	$\begin{bmatrix} 0.99 & 0.02 & -0.00 & 0.01 & -0.09 & -0.01 & 0.13 \\ 0.05 & 0.85 & 0.01 & -0.07 & -0.28 & 0.01 & -0.01 \\ -0.06 & 0.15 & 0.98 & 0.04 & -0.25 & 0.00 & 0.02 \\ -0.10 & -0.06 & -0.03 & 0.88 & 0.13 & -0.09 & -0.27 \\ 0.08 & 0.04 & 0.03 & 0.11 & 1.07 & -0.04 & -0.12 \\ -0.14 & -0.02 & -0.05 & -0.15 & 0.23 & 1.03 & 0.51 \\ -0.13 & 0.22 & -0.04 & 0.02 & -0.07 & 0.03 & 0.87 \end{bmatrix}$
С	$\begin{bmatrix} 0.02 & 0.01 & -0.01 & -0.03 & -0.01 & 0.07 & -0.03 \end{bmatrix}^T$	$\begin{bmatrix} 0.01 & 0.01 & -0.01 & -0.03 & -0.01 & 0.05 & -0.03 \end{bmatrix}^T$
S	$\begin{bmatrix} -0.01 & 0.09 & -0.10 & -0.02 & 0.00 & -0.02 & -0.20 \end{bmatrix}^T$	$\begin{bmatrix} 0.00 & 0.01 & -0.01 & 0.00 & 0.00 & 0.00 & -0.02 \end{bmatrix}^T$
t	0	$\begin{bmatrix} 0.01 & 0.00 & -0.02 & 0.00 & 0.00 & 0.00 \end{bmatrix}^T$
σ	0.851	0.957
δ	_0.003	0.008





the responses at two test points in the region of interest by Sonnet's em (•): the coarse model (---), the enhanced coarse model (—)







the errors of the coarse model responses at the test points







the errors of the enhanced coarse model responses at the test points







The Space Mapping Concept

(Bandler et al., 1994-)







Neural Space Mapping

(*Bandler et al., 1999*)







Space Mapping Based Neuromodeling

(Bandler et. al., 1999)







Neuromappings

Space Mapped neuromapping

Frequency-Dependent Space Mapped neuromapping









Neuromappings (continued)

Frequency Mapped neuromapping

Frequency Space Mapped neuromapping









Neuromappings (continued)

Frequency Partial-Space Mapped neuromapping



it is not always necessary to map the whole set of design parameters

coarse model sensitivities can be used to select the mapped parameters





Training the SM-Based Neuromodel

$$\boldsymbol{w}^* = \arg\min_{\boldsymbol{w}} \left\| \begin{bmatrix} \cdots & \boldsymbol{e}_s^T & \cdots \end{bmatrix}^T \right\|$$
$$\boldsymbol{e}_s = \boldsymbol{R}_f(\boldsymbol{x}_f^{(l)}, \boldsymbol{\omega}_j) - \boldsymbol{R}_c(\boldsymbol{x}_{c_j}^{(l)}, \boldsymbol{\omega}_{c_j}) \qquad \boldsymbol{e}_s \in \Re^r$$
$$\begin{bmatrix} \boldsymbol{x}_{c_j}^{(l)} \\ \boldsymbol{\omega}_{c_j} \end{bmatrix} = \boldsymbol{P}(\boldsymbol{x}_f^{(l)}, \boldsymbol{\omega}_j, \boldsymbol{w})$$
$$j = 1, \dots, F_p \qquad l = 1, \dots, 2n+1 \qquad s = j + F_p(l-1)$$

r is the number of responses in the model

P is the neuromapping function and *w* contains the free parameters of the ANN 2n+1 is the number of training base points and F_p is the number of frequency points Huber optimization is used to solve this problem





Starting Point and Learning Samples

we chose a unit mapping ($\mathbf{x}_c \approx \mathbf{x}_f$ and $\omega_c \approx \omega$) as the starting point for the optimization problem

2n+1 points are used for a microwave circuit with *n* design parameters







Microstrip Right Angle Bend



region of interest $20\text{mil} \le W \le 30\text{mil}$ $8\text{mil} \le H \le 16\text{mil}$ $8 \le \varepsilon_r \le 10$ $1\text{GHz} \le \omega \le 41\text{GHz}$

"coarse" model: equivalent circuit model (Gupta, Garg and Bahl, 1979)

"fine" model: Sonnet's em^{TM}

learning set: 7 base points with "star" distribution





Microstrip Right Angle Bend Coarse Model Errors

comparison between em^{TM} and coarse model at 50 random test points







SM Neuromodel for the Right Angle Bend (3LP:3-6-3)



 $\boldsymbol{x}_{f} = [W \ H \ \varepsilon_{r}]^{T}$





SM Neuromodel Results for the Right Angle Bend

comparison between *em*TM and the SM neuromodel







FDSM Neuromodel for the Right Angle Bend (3LP:4-7-3)



 $\boldsymbol{x}_{f} = [W \ H \ \varepsilon_{r}]^{T}$





FDSM Neuromodel Results for the Right Angle Bend

comparison between *em*TM and the FDSM neuromodel







FSM Neuromodel for the Right Angle Bend (3LP:4-8-4)



 $\boldsymbol{x}_{f} = [W \ H \ \varepsilon_{r}]^{T}$





FSM Neuromodel Results for the Right Angle Bend

comparison between *em*TM and the FSM neuromodel







HTS Quarter-Wave Parallel Coupled-Line Microstrip Filter

(Westinghouse, 1993)



region of interest

```
175 \text{mil} \le L_1 \le 185 \text{mil}
     190 \text{mil} \le L_2 \le 210 \text{mil}
     175mil \leq L_3 \leq 185mil
       18 \text{mil} \le S_1 \le 22 \text{mil}
       75mil \leq S_2 \leq 85mil
       70\text{mil} \le S_3 \le 90\text{mil}
3.901GHz \leq \omega \leq 4.161GHz
```

$$L_0 = 50 \text{mil}$$

$$H = 20 \text{mil}$$

$$W = 7 \text{mil}$$

$$\varepsilon_r = 23.425$$

$$\text{loss tangent} = 3 \times 10^{-5}$$





HTS Microstrip Filter: Fine and Coarse Models

fine model:

Sonnet's em^{TM} with high resolution grid



coarse model:

OSA90/hope[™] built-in models of open circuits, microstrip lines and coupled microstrip lines





HTS Filter Responses Before Neuromodeling

responses using $em^{TM}(\bullet)$ and OSA90/hopeTM(-) at three learning and three test points







HTS Coarse Model Error w.r.t. *em*[™] before any Neuromodeling



in the learning set

in the testing set

learning set: 13 base points with "star" distribution

testing set: 7 random base points in the region of interest (not seen in the learning set)





FM Neuromodel for the HTS Filter (3LP:7-5-1)



 $\boldsymbol{x}_{f} = [L_{1} \ L_{2} \ L_{3} \ S_{1} \ S_{2} \ S_{3}]^{T}$





FM Neuromodel for the HTS Filter (3LP:7-5-1)

responses using $em^{TM}(\bullet)$ and FMN model (-) at the three learning and three testing points







HTS FM Neuromodel Error w.r.t. em[™]



in the learning set

in the testing set





FPSM Neuromodel for the HTS Filter (3LP:7-7-3)



 $\boldsymbol{x}_{f} = [L_{1} \ L_{2} \ L_{3} \ S_{1} \ S_{2} \ S_{3}]^{T} \qquad \boldsymbol{x}_{f}^{\bullet} = [L_{2} \ L_{3} \ S_{2} \ S_{3}]^{T} \qquad \boldsymbol{x}_{c}^{\bullet} = [L_{1c} \ S_{1c}]^{T}$





FPSM Neuromodel for the HTS Filter (3LP:7-7-3)

responses using $em^{TM}(\bullet)$ and FPSMN model (-) at the three learning and three testing points







HTS FPSM Neuromodel Error w.r.t. *em*™



in the learning set

in the testing set





FPSM Neuromodel for the HTS Filter: Fine Frequency Sweep Results

comparison between $em^{TM}(\bullet)$ and FPSMN model (-) at two learning and one testing points







Other Applications of SM based Neuromodels

(Bandler et al., 2000, 2001)

Neural Space Mapping (NSM) Optimization

EM-based Statistical Analysis

EM-based Yield Optimization

Neural Inverse Space Mapping (NISM) Optimization





Conclusions

we describe applications of Space Mapping technology to modeling

we review Generalized Space Mapping (GSM) as an engineering device modeling framework

SM based neuromodeling techniques are also reviewed

frequency-sensitive neuromappings expand the usefulness of empirical quasi-static models

Space Mapping based models can be exploited for efficient EM optimization, statistical analysis and yield optimization





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