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Abstract

This paper presents the theoretical background for designing tests which are topologically sufficient for identification of faulty parameter values in linear subnetworks. Nodal voltages are assumed to be obtainable either by measurements or, indirectly, as a result of a nodal fault analysis. A formulation of nodal fault analysis for subnetworks is presented. It is shown how this approach can be used to evaluate faulty elements within inaccessible faulty subnetworks. The objective of this work is the reduction of the number of required current excitations and, thereby, the number of voltage measurements. Coates flow-graph representation of a network is used.

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I. INTRODUCTION

Fault diagnosis and automatic testing techniques for analog circuits often require parameter identification. Recent papers on the subject [1-16] present different techniques of parameter identification and/or fault region location involving the solution of linear equations. Most of the authors assume voltage measurements, which are more convenient in practice, and consider current excitations only.

A central problem is the formulation of a sufficient number of independent equations subject to a specified number of excitations or voltage measurements. For linear analog circuits, necessary and sufficient conditions related to the network topology have been formulated, resulting in the identification of faulty nodes or subnetworks [12-16].

The principal aim of this paper is to develop topologically based necessary and sufficient conditions for the evaluation of faulty elements within a linear subnetwork under test with a reasonably small number of excitations at a single frequency and, thereby, a small number of measurements. The paper extends the results presented by Biernacki and Starzyk [9] and proposes an efficient approach to the design of test nodes. The Coates flow graph representation of network elements is used [17].

II. LOCATION OF FAULTY NODES AND DESIGN OF NODAL VOLTAGES

Necessary and sufficient conditions for location of faulty nodes have been discussed [14,15] for linear networks, and more generally [13,18] for subnetworks selected during the fault location process in a large network. External voltages and currents of a subnetwork may be

measured or designed through identification of nonfaulty parts of a large network [13].

Consider the nodal equations for a nominal subnetwork isolated during a fault location process for a large network as

$$\tilde{I}^0 = \tilde{Y}^0 \tilde{V}^0. \quad (1)$$

Four types of external nodes are associated with this subnetwork: α -nodes, where both voltages and currents are known; β -nodes, where only voltages are known; γ -nodes, where only currents are known; and δ -nodes, where neither voltages nor currents are known.

We assume that all the elements spanned over the nodes β and δ have been arbitrarily associated with other subnetworks and they are not represented in (1). See Fig. 1.

Solving (1) we obtain

$$\begin{bmatrix} \tilde{V}^{\alpha 0} \\ \tilde{V}^{\beta 0} \\ \tilde{V}^{\gamma 0} \\ \tilde{V}^{\delta 0} \\ \tilde{V}^{\zeta 0} \end{bmatrix} = \begin{bmatrix} \tilde{Z}_{\alpha\alpha} & \tilde{Z}_{\alpha\beta} & \tilde{Z}_{\alpha\gamma} & \tilde{Z}_{\alpha\delta} & \tilde{Z}_{\alpha\zeta} \\ \tilde{Z}_{\beta\alpha} & \tilde{Z}_{\beta\beta} & \tilde{Z}_{\beta\gamma} & \tilde{Z}_{\beta\delta} & \tilde{Z}_{\beta\zeta} \\ \tilde{Z}_{\gamma\alpha} & \tilde{Z}_{\gamma\beta} & \tilde{Z}_{\gamma\gamma} & \tilde{Z}_{\gamma\delta} & \tilde{Z}_{\gamma\zeta} \\ \tilde{Z}_{\delta\alpha} & \tilde{Z}_{\delta\beta} & \tilde{Z}_{\delta\gamma} & \tilde{Z}_{\delta\delta} & \tilde{Z}_{\delta\zeta} \\ \tilde{Z}_{\zeta\alpha} & \tilde{Z}_{\zeta\beta} & \tilde{Z}_{\zeta\gamma} & \tilde{Z}_{\zeta\delta} & \tilde{Z}_{\zeta\zeta} \end{bmatrix} \begin{bmatrix} \tilde{I}^{\alpha 0} \\ \tilde{I}^{\beta 0} \\ \tilde{I}^{\gamma 0} \\ \tilde{I}^{\delta 0} \\ 0 \end{bmatrix}, \quad (2)$$

where ζ represents internal nodes and \tilde{Z}_{ab} denotes a submatrix of $(\tilde{Y}^0)^{-1}$ obtained by the intersection of rows a and columns b .

For any subnetwork, with $\text{card } \alpha > \text{card } \delta$, we obtain an internal-self-testing condition [18]:

$$\begin{bmatrix} \tilde{V}^{\alpha} \\ \tilde{V}^{\beta} \end{bmatrix} - \begin{bmatrix} \tilde{Z}_{\alpha\alpha} & \tilde{Z}_{\alpha\gamma} \\ \tilde{Z}_{\beta\alpha} & \tilde{Z}_{\beta\gamma} \end{bmatrix} \begin{bmatrix} \tilde{I}^{\alpha} \\ \tilde{I}^{\gamma} \end{bmatrix} = \begin{bmatrix} \tilde{Z}_{\alpha\beta} & \tilde{Z}_{\alpha\delta} \\ \tilde{Z}_{\beta\beta} & \tilde{Z}_{\beta\delta} \end{bmatrix} \begin{bmatrix} \tilde{I}^{\beta} \\ \tilde{I}^{\delta} \end{bmatrix}. \quad (3)$$

Let $\tilde{Z}_{a_1 \dots a_k, b_1 \dots b_m}$ denote a submatrix of $(\tilde{Y}^0)^{-1}$ obtained by the intersection of rows a_1 u a_2 u ... u a_k and columns b_1 u b_2 u ... u b_m .

Result 1 (Fault-free subnetworks)

If the system of equations (3) is consistent and

$$\text{Rank}[Z_{\alpha\beta, \beta\delta x}] > \text{Rank}[Z_{\alpha\beta, \beta\delta}], \quad (4)$$

where $x \in \alpha \cup \gamma \cup \zeta$, then there are no faulty elements incident with nodes x .

According to Result 1, only the elements spanned over the external nodes $\beta \cup \delta$ can be faulty. Because we have associated these elements with other subnetworks we can declare the subnetwork under consideration as fault free. Equation (3) can then be solved for I_{α}^{β} and I_{α}^{δ} , hence all the voltages of this subnetwork can be calculated. Consequently, the β - and δ -nodes of this subnetwork become α -nodes of adjacent subnetworks.

Let nodes $\eta \in \alpha \cup \gamma \cup \zeta$ be faulty, and $\text{card } \alpha > (\text{card } \delta) + (\text{card } \eta)$. Let

$$V^{\alpha\beta} \triangleq \begin{bmatrix} V^{\alpha} \\ \tilde{V}^{\beta} \\ V^{\beta} \end{bmatrix}, \quad I^{\alpha\gamma} \triangleq \begin{bmatrix} I^{\alpha} \\ \tilde{I}^{\gamma} \\ I^{\gamma} \end{bmatrix}, \quad I^{\beta\delta} \triangleq \begin{bmatrix} I^{\beta} \\ \tilde{I}^{\delta} \\ I^{\delta} \end{bmatrix}$$

and let I^{η} be the vector of node currents representing faults.

Result 2 [18] (Faulty subnetworks)

If the system of equations

$$V^{\alpha\beta} - Z_{\alpha\beta, \alpha\gamma} I^{\alpha\gamma} = Z_{\alpha\beta, \beta\delta} I^{\beta\delta} + Z_{\alpha\beta, \eta} I^{\eta} \quad (6)$$

is consistent and

$$\text{Rank}[Z_{\alpha\beta, \beta\delta\eta x}] > \text{Rank}[Z_{\alpha\beta, \beta\delta}] + \text{card } \eta, \quad (7)$$

where $x \in \alpha \cup \gamma \cup \zeta - \eta$, then the only faulty elements can be those spanned over the set of nodes $F = \eta \cup \beta \cup \delta$. These nodes are called faulty nodes although there may be no faulty element incident with β and δ .

Assume that by solving (6) we have evaluated I_{α}^{β} , I_{α}^{δ} and I_{α}^{η} . We can again proceed to evaluate all voltages of the subnetwork under consideration and use the information obtained to analyse the adjacent subnetworks.

Let \tilde{T}^a denote a matrix obtained from the unity matrix of the same dimensions as the \tilde{Y}^0 matrix by extracting columns which correspond to elements a. $\tilde{T}^a \tilde{I}^a$ will transform vector \tilde{I}^a into a vector with the dimension of \tilde{Y}^0 , where nonzero elements can only appear in rows corresponding to the elements of \tilde{I}^a . For example,

$$\tilde{T}^\alpha \tilde{I}^\alpha = \begin{bmatrix} \tilde{I}^\alpha \\ 0 \end{bmatrix}.$$

For the assumed faulty subnetwork, (1) can be replaced by

$$\tilde{I} = \tilde{Y} \tilde{V}, \quad (8)$$

where

$$\begin{aligned} \tilde{I} &\triangleq \tilde{I}^0 + \Delta \tilde{I} = \\ &= \tilde{T}^\alpha \tilde{I}^{\alpha 0} + \tilde{T}^\gamma \tilde{I}^{\gamma 0} + \tilde{T}^F \tilde{I}^F, \end{aligned} \quad (9)$$

$$\tilde{I}^F = \begin{bmatrix} \tilde{I}^\eta \\ \tilde{I}^{\beta\delta} \end{bmatrix} \quad (10)$$

and

$$\tilde{V} = \tilde{V}^0 + \Delta \tilde{V} \quad (11)$$

is the vector of nodal voltages in the faulty network. After solving (6), we know the left-hand side of (8) and we can solve (8) to get \tilde{V} .

For all independent current excitations we are, therefore, able to calculate voltages in the faulty network if the conditions of Result 2 are satisfied. These voltages, which would otherwise have to be measured, are required by the approach presented in [9] for evaluating all the elements of a network. In the present paper we only need to evaluate unknown elements, i.e., those which are spanned over the faulty nodes.

III. ELEMENT EVALUATION FOR SUBNETWORKS SPANNED OVER FAULTY NODES

The elements spanned over faulty nodes may form separate subnetworks within a given subnetwork, as shown in Fig. 1. The subnetworks may be remote and inaccessible from the point of view of direct excitation and measurement. We can formulate conditions for element evaluation within each of these subnetworks separately and combine the results obtained to establish conditions for the whole network. These conditions will show which external nodes should be excited independently to evaluate all faulty elements.

Consider a linear subnetwork spanned over N faulty nodes. Let the N -dimensional vectors \underline{I}_s and \underline{V}_s be subsets of \underline{I} and \underline{V} , respectively, corresponding to this subnetwork. We can then write

$$\underline{Y}_s \underline{V}_s^i = \underline{I}_s^i, \quad (12)$$

for the i th excitation. Our goal is to evaluate \underline{Y}_s and then the element values. Although we concentrate our discussion on the nodal equations, it is applicable to any other description based on an independent set of cut-sets (see [9]).

For N independent excitations, we can write a matrix equation

$$\underline{Y}_s \underline{V}_t = \underline{I}_t, \quad (13)$$

where the square matrix

$$\underline{V}_t \triangleq [\underline{V}_s^1 \quad \underline{V}_s^2 \quad \dots \quad \underline{V}_s^N] \quad (14)$$

is the matrix of voltage responses and the square matrix

$$\underline{I}_t \triangleq [\underline{I}_s^1 \quad \underline{I}_s^2 \quad \dots \quad \underline{I}_s^N] \quad (15)$$

is the matrix of current excitations. From (13), we find the unknown matrix \underline{Y}_s as

$$\underline{Y}_s = \underline{I}_t \underline{V}_t^{-1} \quad (16)$$

provided that \underline{V}_t is nonsingular. As a consequence of equations (13) and

(16), the following result provides sufficient conditions for the evaluation of \underline{Y}_s .

Result 3 [9]

If a given linear subnetwork can be described by the nodal equation (12) and the current excitations are chosen in such a way that \underline{I}_t is a nonsingular matrix, then \underline{V}_t is also nonsingular and the solution (16) exists.

Proof of this result follows from equation (12) since $N = \text{rank } \underline{I}_t \leq \text{rank } \underline{V}_t \leq N$.

Thus, in order to identify the values of all elements of \underline{Y}_s , we could arrange for N independent current excitations, design or measure all nodal voltages and then apply equation (16).

In order to perform the least number of tests, however, we must obviously eliminate whole columns of \underline{V}_t . We propose a systematic way which enables us to identify tests sufficient for component evaluation. The method assumes that all components have nonzero values.

Conditions for Sufficient Tests

Equation (13) can be rewritten in the form

$$\underline{V}_t^T \underline{Y}_s^T = \underline{I}_t^T \tag{17}$$

Consider the product of \underline{V}_t^T and the jth column of \underline{Y}_s^T . We have

$$\underline{V}_t^T \underline{y}_j = \begin{bmatrix} \underline{V}_s^{1T} \\ \underline{V}_s^{2T} \\ \cdot \\ \cdot \\ \underline{V}_s^{NT} \end{bmatrix} \begin{bmatrix} y_{j1} \\ y_{j2} \\ \cdot \\ \cdot \\ y_{jN} \end{bmatrix} = \begin{bmatrix} I_{sj}^1 \\ I_{sj}^2 \\ \cdot \\ \cdot \\ I_{sj}^N \end{bmatrix} , \tag{18}$$

where I_{sj}^i is the j th element of the vector \tilde{I}_s^i .

Let the k unknown elements of \tilde{y}_j be identified by the set of indices $B_j = \{j_1, \dots, j_k\}$. We denote the set of elements y_{ji} , $i \in B_j$ a reduced cut-set. Transferring the known terms from the left-hand side to the right-hand side of (18) and adjusting I_{sj}^i appropriately we rewrite the equation as

$$\tilde{V}_t^T \tilde{y}_{B_j} = \begin{bmatrix} \tilde{V}_s^{1T} \\ \tilde{V}_s^{2T} \\ \vdots \\ \tilde{V}_s^{NT} \end{bmatrix} \begin{bmatrix} 0 \\ \vdots \\ 0 \\ y_{jj_1} \\ \vdots \\ y_{jj_k} \\ \vdots \\ 0 \end{bmatrix} = \begin{bmatrix} I_{1j} \\ I_{2j} \\ \vdots \\ I_{Nj} \end{bmatrix}, \quad (19)$$

where I_{ij} is equivalent external current for a reduced cut-set at the j th node due to the i th current excitation.

In order to determine the elements $y_{jj_1}, \dots, y_{jj_k}$, we can solve a subsystem of (19) given by

$$\tilde{V}_t^T [C_j | B_j] \begin{bmatrix} y_{jj_1} \\ \vdots \\ \vdots \\ y_{jj_k} \end{bmatrix} = \begin{bmatrix} I_{i_1j} \\ \vdots \\ \vdots \\ I_{i_kj} \end{bmatrix}, \quad (20)$$

where the k equations are chosen from (19) in such a way that the square submatrix $\tilde{V}_t^T [C_j | B_j]$, obtained as the intersection of rows $C_j = \{i_1,$

..., i_k and columns B_j , is nonsingular. See Fig. 2 for an illustration. According to relationship (13), the matrix $\underset{\sim}{V}_t^T [C_j | B_j]$ can be defined as

$$\underset{\sim}{V}_t^T [C_j | B_j] = \underset{\sim}{I}_t^T [C_j | \cdot] (\underset{\sim}{Y}_s^T)^{-1} [\cdot | B_j], \quad (21)$$

where $\underset{\sim}{I}_t^T [C_j | \cdot]$ consists of rows C_j from $\underset{\sim}{I}_t^T$ and $(\underset{\sim}{Y}_s^T)^{-1} [\cdot | B_j]$ consists of columns B_j from $(\underset{\sim}{Y}_s^T)^{-1}$. On the basis of (21) and the Cauchy-Binet theorem [19] we may formulate the following result.

Result 4

The matrix $\underset{\sim}{V}_t^T [C_j | B_j]$ is nonsingular if and only if

$$\exists A_j \quad \det \underset{\sim}{I}_t^T [C_j | A_j] \neq 0 \text{ and } \det \underset{\sim}{Y}_s (A_j | B_j) \neq 0, \quad (22)$$

where $\underset{\sim}{Y}_s (A_j | B_j)$ denotes the submatrix of $\underset{\sim}{Y}_s$ obtained by removing rows A_j and columns B_j (see Fig. 2).

Consider a sequence of sets B_j , $j = j_1, \dots, j_M$, which corresponds to a sequence of reduced cut-sets of the current graph of the subnetwork. Only those reduced cut-sets will be considered for which external currents, if any, can be specified. Based on (13) and (20), the following result can be summarized.

Result 5

Independent excitations which appear at or are applied to the subset of nodes $A \subset \{1, 2, \dots, N\}$ are sufficient for the identification of all elements of $\underset{\sim}{Y}_s$ if and only if

$$\forall B_j \quad \exists C_j \quad \exists A_j \subset A, \det \underset{\sim}{I}_t^T [C_j | A_j] \neq 0 \text{ and } \det \underset{\sim}{Y}_s (A_j | B_j) \neq 0, \quad (23)$$

where

$$\text{card } A_j = \text{card } B_j = \text{card } C_j. \quad (24)$$

Nodes A in Result 5 can be chosen from a remote inaccessible subnetwork therefore we call them injection nodes. For each subnetwork the set A must be a subset of the external nodes of this subnetwork.

As a consequence of (24), we have the following corollary.

Corollary 1

$$\text{card } A \geq \max_j \text{card } B_j . \quad (25)$$

It is seen from (25) that the choice of the sequence of B_j is crucial for the minimization of the number of sufficient tests.

Now, in order to characterize A_j feasible for a given B_j , we consider topological equations for the nodal admittance matrix.

$$\underline{Y}_s = \underline{\lambda}_- \underline{Y}_e \underline{\lambda}_+^T , \quad (26)$$

where the element ij of $\underline{\lambda}_-$ is equal to 1 if the j th edge is directed towards the i th node, otherwise zero; and the element ij of $\underline{\lambda}_+$ is equal to 1 if the j th edge is directed away from the i th node, otherwise zero; \underline{Y}_e is a diagonal matrix of element admittances.

The submatrix of \underline{Y}_s obtained by removing columns B_j can be expressed as

$$\underline{Y}_s(\cdot \mid B_j) = \underline{\lambda}_- \underline{Y}_e \underline{\lambda}_+^{\prime T} , \quad (27)$$

where $\underline{\lambda}_+^{\prime}$ is obtained from $\underline{\lambda}_+$ by removing rows B_j . In the Coates graph, this corresponds to deleting all the edges outgoing from nodes B_j .

Similarly,

$$\underline{Y}_s(A_j \mid \cdot) = \underline{\lambda}_-^{\prime} \underline{Y}_e \underline{\lambda}_+^{\prime T} , \quad (28)$$

where $\underline{\lambda}_-^{\prime}$ is obtained from $\underline{\lambda}_-$ by removing rows A_j . In the Coates graph, this corresponds to deleting all the edges incoming to nodes A_j .

Let G denote a directed Coates graph [17] and let P denote a set of node pairs of G , namely, $P = \{(v_{s1}, v_{e1}), \dots, (v_{sk}, v_{ek})\}$, where $v_{p\ell} \neq v_{nm}$ for $\ell \neq m$ ($p, n = s, e$).

Definition [20]

A k -connection of a graph G is a subgraph c_p of the graph, such that elements of c_p form a set of k node-disjoint directed paths and

node-disjoint directed circuits incident with all graph nodes. The starting point and the endpoint of the paths are indicated by the pairs of P .

Let us consider the Coates graph $G(A_j | B_j)$ obtained from the graph of the given subnetwork after deleting all the edges incoming to nodes A_j and all the edges outgoing from nodes B_j . The following theorem can be proved on the basis of the Cauchy-Binet theorem [19] and the concept of the k -connection [20].

Theorem 1

If $\det \underline{Y}_S(A_j | B_j) \neq 0$, there exists in $G(A_j | B_j)$ at least one k -connection c_p (see Fig. 3), where

$$P = \{(v_s, v_e) | v_s \in A_j, v_e \in B_j\} \quad (29)$$

and

$$k = \text{card } P = \text{card } A_j = \text{card } B_j . \quad (30)$$

(v_s, v_e) represents a path directed from the node v_s to the node v_e or isolated node when $v_s = v_e$, and N denotes all nodes of the graph. The condition stated in the theorem is sufficient almost everywhere, which means that if a specified k -connection c_p exists then $\det \underline{Y}_S(A_j | B_j) \neq 0$ for almost all values of element admittances.

Proof

According to the Cauchy-Binet theorem and relation (28), we have

$$\det \underline{Y}_S(A_j | B_j) = \Sigma \det \underline{C}^- \cdot \det \underline{C}^+ , \quad (31)$$

where \underline{C}^- is a major submatrix of $\underline{\lambda}'_- \cdot \underline{Y}_e$ with order equal to $(N - \text{card } A_j)$ and \underline{C}^+ is the corresponding major submatrix of $\underline{\lambda}'_+{}^T$. If $\det \underline{Y}_S(A_j | B_j) \neq 0$, then there exists at least one pair of corresponding determinants, both different from zero. A major determinant of $\underline{\lambda}'_- \cdot \underline{Y}_e$ is different from zero if and only if there exists one nonzero element in every row

of the chosen submatrix (chosen set of columns). This corresponds to the set of $(N - \text{card } A_j)$ edges, such that every edge has a different endpoint, belonging to the set of nodes $(N - A_j)$. The corresponding submatrix is different from zero if the same edges have different origins, belonging to the same set of nodes $(N - B_j)$. Now it is easy to check that these edges form a k -connection, as stated in Theorem 1. The determinant of $\underline{Y}_s(A_j \mid B_j)$ equals zero, in spite of having nonzero components in (31), only for particular element values. So, if a specified k -connection c_p exists then $\underline{Y}_s(A_j \mid B_j) \neq 0$ with conditional probability equal 1.

Remark

If $\text{rank}(\underline{I}_t^T [\cdot \mid A]) = \text{card } A$, where $\underline{I}_t^T [\cdot \mid A]$ consists of columns A from \underline{I}_t^T , then

$$\forall A_j \subset A \exists c_j \det \underline{I}_t^T [c_j \mid A_j] \neq 0 . \quad (32)$$

As a consequence of Theorem 1 and the Remark, we have an important corollary.

Corollary 2

To satisfy (23), we should find a set A_j such that, after deleting all the edges outgoing from nodes B_j and after deleting all the edges incoming to nodes A_j , there are no isolated nodes in the set $N - (A_j \cup B_j)$.

Theorem 1 does not guarantee that $\underline{Y}_s(A_j \mid B_j)$ is nonsingular. It may, however, be singular for particular element values only. If we know the nominal values of the elements, then we can easily check whether the injection nodes A_j are sufficient for the solution.

Definition

A node is said to be a corner if there exists a complete subgraph

containing all the edges incoming to the node as well as the edges having the same weight as any of the incoming ones.

It follows that there may exist edges outgoing from a corner to other parts of the graph. Also, the order of the complete graph is not defined. In particular, it may be a complete graph of zero order (see Fig. 4a).

Based on Corollary 2, the following theorem can be proved.

Theorem 2

All the corners must be injection nodes.

Proof

Assume that a corner is not an injection node. If we identify an edge within the subnetwork incident with the corner, then every reduced cut-set containing the edge must contain all the nodes of the complete subgraph. After deleting all the edges outgoing from the nodes of this reduced cut-set, the corner will be an isolated node, and if it is not an injection node, we obtain an isolated node in the set $N - (A_j \cup B_j)$ and a contradiction to Corollary 2.

Thus, the number of corners influences the minimal cardinality of A. In order to estimate the cardinality of A, the following remarks may be helpful.

Remark 1 $\text{card } A \geq \text{order of the maximal complete subgraph.}$

Remark 2 $\text{card } A \geq \text{minimal incoming degree in the remaining subgraph}$
after deleting all edges incident with corners.

The incoming degree of a vertex is the number of edges incoming to this vertex.

Location of Injection Nodes

An optimal selection of injection nodes could be done in a combina-

torial way, where different sets of reduced cut-sets are considered and then different combinations of injection nodes are checked. However, for large networks, it may be quite tedious to check the conditions of Theorem 1, even if reduced cut-sets and a set A are known.

A very efficient heuristic algorithm which utilizes theoretical aspects discussed at this section was presented in [9]. It allows us to find a nearly minimal set of injection nodes in a time which depends linearly on the subnetwork size. Since the conditions stated in Theorem 1 must be satisfied, the algorithm localizes injection nodes in such a way that there exist a set of separate paths from injection nodes to the nodes of each reduced cut-set, as illustrated in Fig. 5.

In particular cases, when the number of injection nodes is too large because of the subnetwork topology we can reduce them by adding some known elements to the subnetwork under consideration. The same argument holds when we have too many corners in the subnetwork (Fig. 6). These remarks concern the case when we identify elements of a given network using voltage measurements at all nodes [9] as well as evaluation of faulty elements within remote, inaccessible subnetworks. In the latter case, adding the known elements may be equivalent to considering an augmented subnetwork which will contain faulty nodes as well as some nonfaulty ones.

The following examples explain how to use the results obtained from the test finding algorithm to identify all network elements.

Example 1

The subnetwork whose parameters we want to design and its Coates graph are shown in Fig. 7 (node 0 is chosen as the reference node). Let us assume for simplicity that the independent current excitation $\underline{I}_t = \underline{1}$.

This can be easily achieved when elements are identified through direct voltage measurements. There are 3 corners in this network - nodes 1, 6 and 7. Using the algorithm presented in [9] we find that they constitute a sufficient set of injection nodes for this network. Table I illustrates the reduced cut-sets considered and elements associated with them. For identification of network elements, we apply excitations at nodes 1, 6 and 7. The nodal voltages measured with unit excitations at different nodes are shown in Table II. We formulate equations (20) for successive reduced cut-sets and compute element values. The first equation is as follows:

$$\begin{bmatrix} V_{11} & V_{12} \\ V_{61} & V_{62} \end{bmatrix} \begin{bmatrix} Y_1 + Y_2 \\ -Y_2 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix} \quad \text{or} \quad \begin{bmatrix} 0.77641 & 0.32925 \\ -0.38775 & -1.1633 \end{bmatrix} \begin{bmatrix} Y_1 + Y_2 \\ -Y_2 \end{bmatrix} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

and we obtain $Y_1 = 1$, $Y_2 = 0.5$.

The second equation

$$\begin{bmatrix} V_{11} & V_{12} & V_{13} & V_{14} \\ V_{61} & V_{62} & V_{63} & V_{64} \\ V_{71} & V_{72} & V_{73} & V_{74} \end{bmatrix} \begin{bmatrix} -Y_2 \\ Y_2 + Y_3 + Y_4 + Y_5 \\ -Y_5 \\ -Y_4 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$$

can be transformed, because Y_2 is now known, to

$$\begin{bmatrix} V_{12} & V_{13} & V_{14} \\ V_{62} & V_{63} & V_{64} \\ V_{72} & V_{73} & V_{74} \end{bmatrix} \begin{bmatrix} Y_3 + Y_4 + Y_5 \\ -Y_5 \\ -Y_4 \end{bmatrix} = \begin{bmatrix} (V_{11} - V_{12}) Y_2 \\ (V_{61} - V_{62}) Y_2 \\ (V_{71} - V_{72}) Y_2 \end{bmatrix}$$

or

$$\begin{bmatrix} 0.32925 & -0.006477 & 0.14264 \\ -1.1633 & -4.6575 & -1.4699 \\ 0.048524 & 0.17534 & 0.076466 \end{bmatrix} \begin{bmatrix} Y_3 + Y_4 + Y_5 \\ -Y_5 \\ -Y_4 \end{bmatrix} = \begin{bmatrix} 0.22358 \\ 0.38778 \\ -0.016175 \end{bmatrix}$$

and we obtain $Y_3 = 0.333$, $Y_4 = 0.25$, $Y_5 = 0.2$.

Continuing the procedure we design all the other network elements as

$$Y_6 = 0.167, Y_7 = 0.143, Y_8 = 0.125, Y_9 = 0.111, Y_{10} = 0.1,$$

$$Y_{11} = 0.0909, Y_{12} = 0.0833, Y_{13} = 0.0769, Y_{14} = 0.0714,$$

$$Y_{15} = 0.0667, Y_{16} = 0.0625, Y_{17} = 0.0588, g_m = 8.5.$$

Example 2

We apply the algorithm proposed to the passive grid circuit shown in Fig. 8. In such circuits, the number of nodes $n = k^2$ and number of passive elements $e = 2k^2 - 2k$, where $k = 2, 3, \dots$. We assume that the voltage at each node is known. Using the algorithm described in [9] we find that no matter what the size of the grids three tests at a single frequency are sufficient for determining all the element values.

IV. ELEMENT EVALUATION USING EXTERNAL EXCITATION NODES

Let us assume that we have distinct, remote, inaccessible faulty subnetworks S_1, \dots, S_f spanned over faulty nodes within the subnetwork under investigation (see Fig. 9). According to Result 2, the number of external nodes, where both voltages and external currents are known, have to satisfy the relation

$$\text{card } \alpha > \sum_{i=1}^f n_i, \quad (33)$$

where n_i is the number of nodes in the subnetwork S_i . We can apply the approach discussed in Section III to each subnetwork S_1, \dots, S_f separately to identify sets of injection nodes A^1, \dots, A^f at which independent current excitations could be forced. With independent

excitations appearing at injection nodes, we are able to evaluate all elements within S_1, \dots, S_f .

Let T be a subset of the external nodes of the subnetwork S , which is defined by (2). Let G denote the Coates signal-flow graph of S . Let us assume that we have evaluated faulty currents and designed nodal voltages as discussed in Section II. Let $k_i = \text{card } A^i$.

Lemma 1

If there exist k_i simultaneous and separate paths in G from T to A^i not incident with other S_i nodes, then all the elements of S_i can be uniquely identified.

Proof is based on the recognition of each cut-set in S_i as a reduced cut-set in S .

Corollary 3

If the Lemma 1 is satisfied for all A^i , then T can be chosen as a set of test nodes where independent current excitations are applied, to evaluate all faulty elements in S .

We are interested to have the cardinality of T as small as possible, to minimize the number of tests and designs of nodal voltages.

Corollary 4

$$\text{card } T \geq \max k_i. \quad (34)$$

The main goal of the approach presented is to find k_i as small as possible, so the technique described guarantees identification of faulty elements effectively. For most practical cases, card T is between 2 and 5.

Remark

For identification of faulty elements within remote inaccessible subnetworks we design currents flowing into these subnetworks from the

surrounding network using the designed voltages and nominal element values first, and then proceed with element evaluation within each of them as discussed.

Example 3

Assume that the nominal element values for the network from Fig. 7 are as follows:

$$\begin{aligned} Y_1 &= 1, Y_2 = 0.5, Y_3 = 0.3, Y_4 = 0.32, \\ Y_5 &= 0.2, Y_6 = 0.167, Y_7 = 0.143, \\ Y_8 &= 0.125, Y_9 = 0.1, Y_{10} = 0.2, Y_{11} = 0.1, \\ Y_{12} &= 0.0833, Y_{13} = 0.0769, Y_{14} = 0.0714, \\ Y_{15} &= 0.0667, Y_{16} = 0.0625, Y_{17} = 0.0588, g_m = 8.5. \end{aligned}$$

Four external points are available for voltage measurements and current excitations at the nodes 1, 3, 4 and 7. Assume for simplicity that all external nodes are of the α type. Using the approach discussed in Section II we have found three faulty nodes, namely, 2, 4, 6 and evaluated currents \underline{I}^n , $n = \{2, 4, 6\}$. The subnetwork spanned over the faulty nodes is a simple ladder network. With the help of the method discussed in Section III we can easily locate nodes 2 and 6 as injection nodes sufficient for evaluation of the ladder elements. According to Lemma 1 external current excitations sufficient for element evaluation can be made at nodes 1 and 7.

Now we simulate the nominal network with independent (unit) excitations at nodes 1 and 7 separately and evaluate currents \underline{I}^n from equation (6). With those currents and independent current excitations we excite the nominal network to obtain the current voltages as in rows 1 and 3 of the Table II. Elements Y_2, Y_5, Y_7 and Y_8 are nominal as they are not spanned over the faulty nodes. Using the voltages from Table II

we calculate external currents for the ladder subnetwork spanned over faulty nodes as equal to

$$I_{12} = (V_{11} - V_{12}) Y_2 + (V_{13} - V_{12}) Y_5 = 0.1564,$$

$$I_{14} = (V_{13} - V_{14}) Y_7 = -0.02135,$$

$$I_{16} = (V_{13} - V_{16}) Y_8 = -0.005633.$$

Similarly we can get

$$I_{72} = 0.009188, I_{74} = 0.01414, I_{76} = 0.01049.$$

Equation (20) for the first reduced cut-set has the form

$$\begin{bmatrix} V_{12} & V_{14} \\ V_{72} & V_{74} \end{bmatrix} \begin{bmatrix} Y_3 + Y_4 \\ -Y_4 \end{bmatrix} = \begin{bmatrix} I_{12} \\ I_{72} \end{bmatrix}$$

Therefore,

$$\begin{bmatrix} 0.32925 & 0.14264 \\ 0.048524 & 0.076466 \end{bmatrix} \begin{bmatrix} Y_3 + Y_4 \\ -Y_4 \end{bmatrix} = \begin{bmatrix} 0.1564 \\ 0.009188 \end{bmatrix}$$

and we get $Y_3 = 0.333$ and $Y_4 = 0.25$. In the next two reduced cut-sets elements Y_9 , Y_{11} and Y_{10} are evaluated respectively, with the help of a voltage measurement as well as evaluated and nominal elements.

V. CONCLUSIONS

The method presented enables us to find a reasonably small number of excitation nodes which are topologically sufficient for the identification of all faulty parameter values of linear analog subnetworks. This can be achieved by searching for a "good" sequence of reduced cut-sets within the subnetworks spanned over faulty nodes, whose elements are consecutively determined from (20). The element evaluation approach as presented in Sections III and IV is easy to program and gives a linear dependence of computational effort on the size of the network. The notion of corner is particularly important, since it

influences the number of necessary injection nodes independently of a sequence of cut-sets. The number of excitations can be reduced by adding external elements or some nominal ones in the case of inaccessible subnetworks.

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TABLE I
REDUCED CUT-SETS

| Step i | Nodes in reduced cut-set | Elements in the reduced cut-set to be found |
|--------|--------------------------|---|
| 1 | 1,2 | Y_1, Y_2 |
| 2 | 2,3,4 | Y_3, Y_4, Y_5 |
| 3 | 3,4,6 | Y_7, Y_9, Y_{11} |
| 4 | 3,6 | Y_8, Y_{10} |
| 5 | 3,5 | Y_6 |
| 6 | 5,7,8 | Y_{12}, Y_{13}, Y_{14} |
| 7 | 7,8 | Y_{15}, Y_{17} |
| 8 | 6,8 | Y_{16}, g_m |

TABLE II
NODAL VOLTAGES FOR EXAMPLE 1

| | | Voltage at node no. | | | | | | | |
|----------------------------|--|---------------------|---------|-----------|---------|---------|---------|---------|---------|
| Excitation at the node no. | | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 |
| 1 | | .77641 | .32925 | -.0066477 | .14264 | -.57149 | .038418 | -.91631 | -2.0943 |
| 6 | | -.38775 | -1.1633 | -4.6575 | -1.4699 | -15.751 | .89959 | -22.757 | -50.343 |
| 7 | | .016174 | .048524 | .17534 | .076466 | .47525 | .091385 | 4.5309 | -2.126 |

FIGURE CAPTIONS

- Fig. 1 Illustration of remote, inaccessible faulty subnetworks (shaded) spanned over faulty nodes.
- Fig. 2 Illustrations of equation (20) and Result 4.
- Fig. 3 Example of required 3-connections.
- Fig. 4 Examples of corners. Corners are denoted by v .
- Fig. 5 Illustration of the paths required.
- Fig. 6 External path from an injection node to a corner.
- Fig. 7 (a) Faulty network, (b) Coates graph.
- Fig. 8 Grid circuit.
- Fig. 9 Inaccessible faulty subnetworks (shaded).

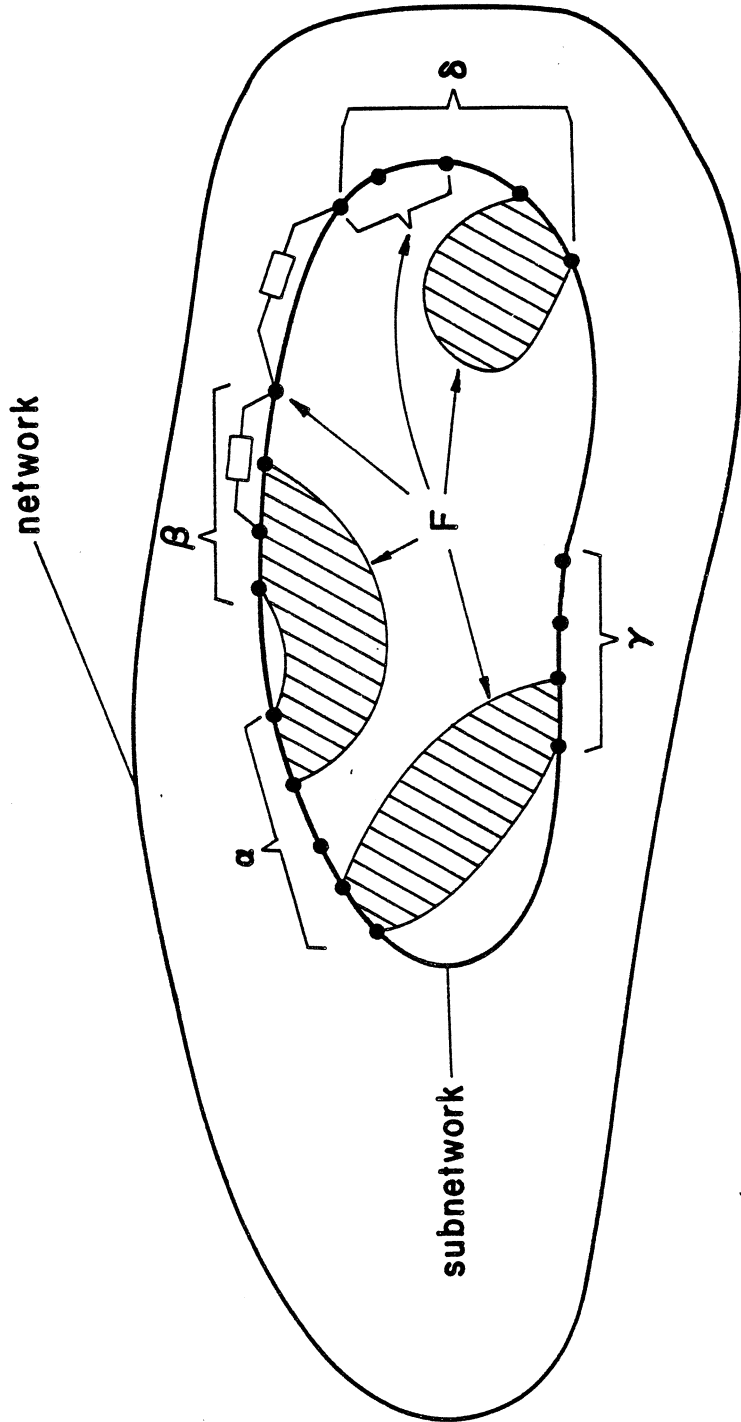


Fig. 1

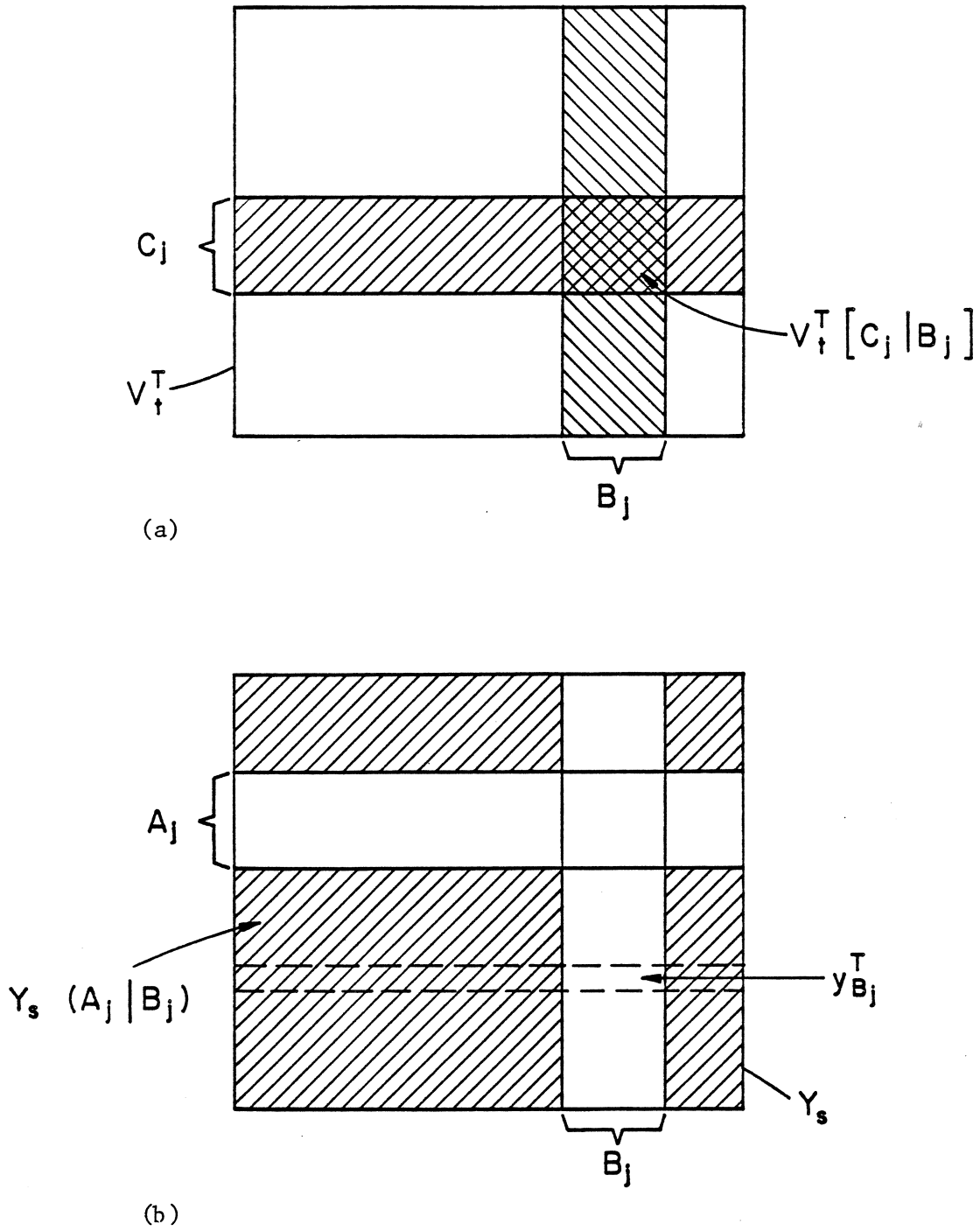


Fig. 2

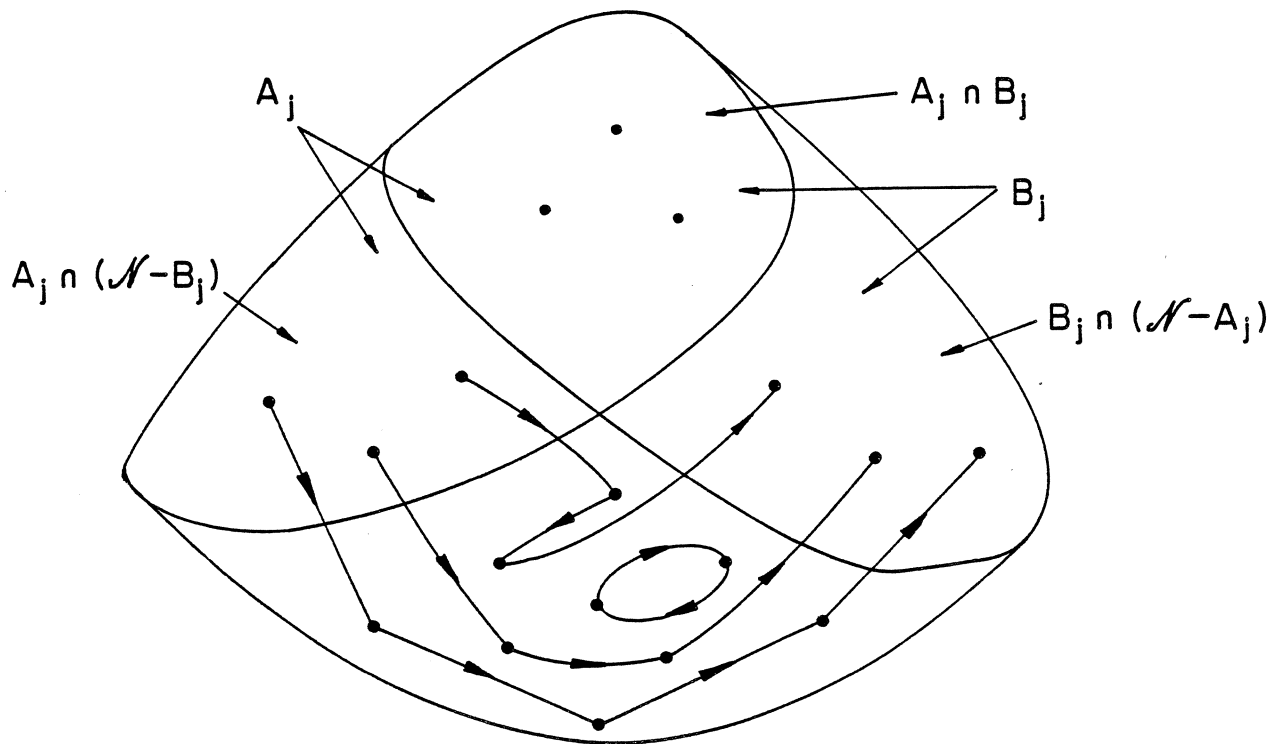
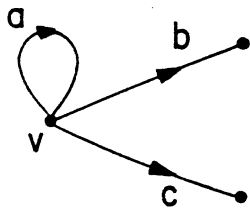
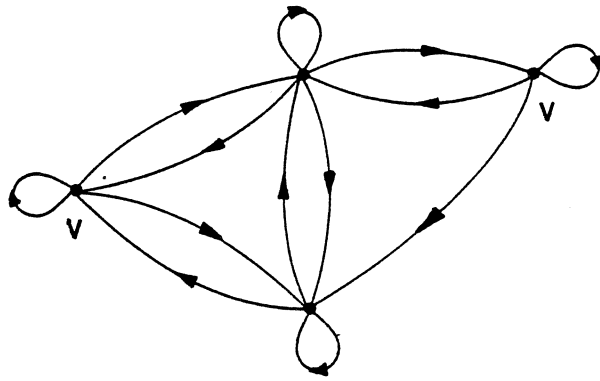


Fig. 3



(a)



(b)

Fig. 4

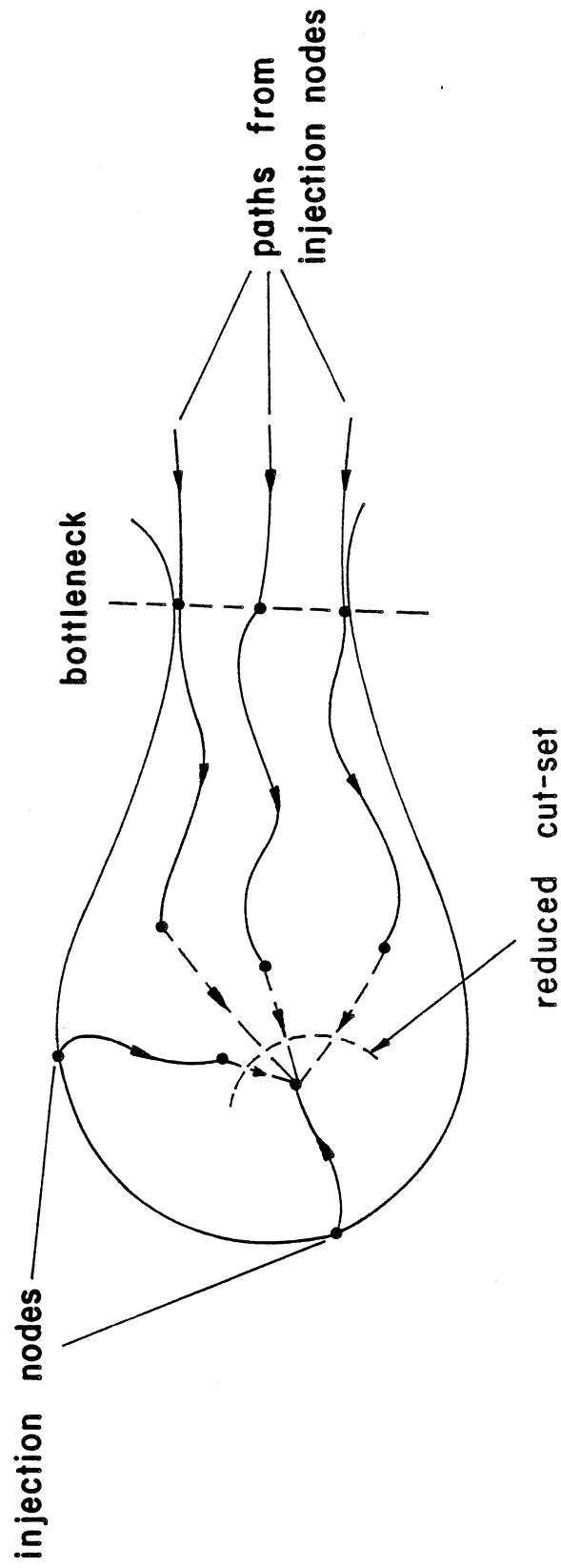


Fig. 5

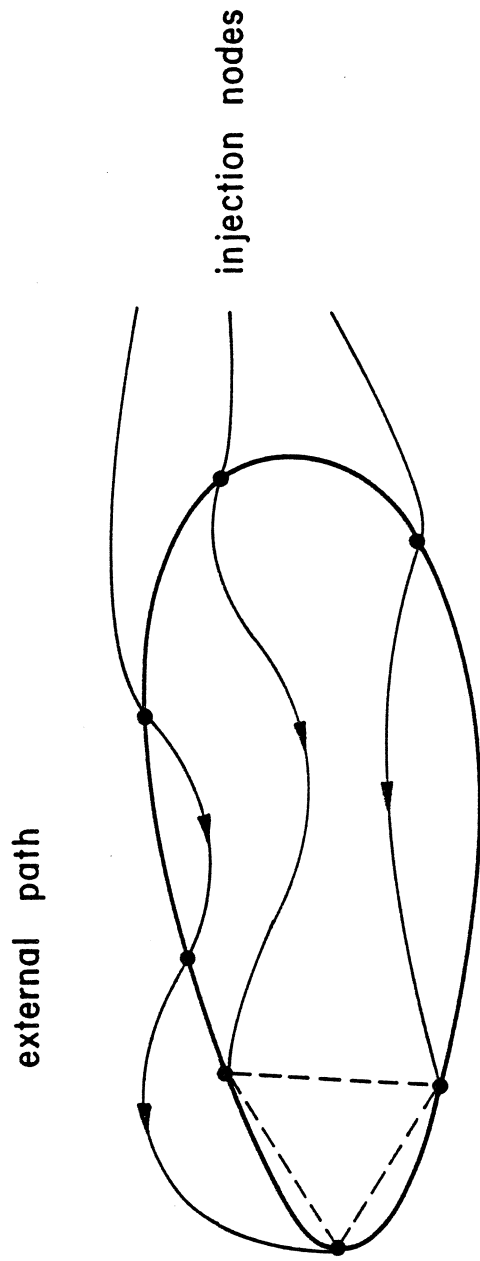


Fig. 6

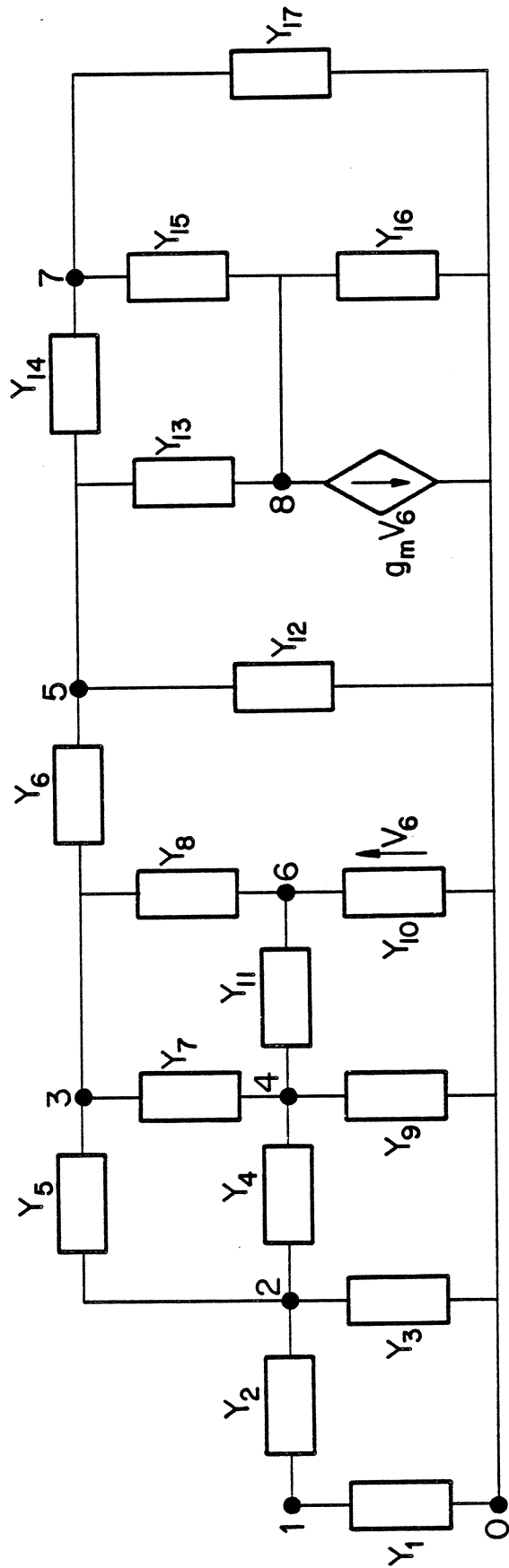


Fig. 7(a)

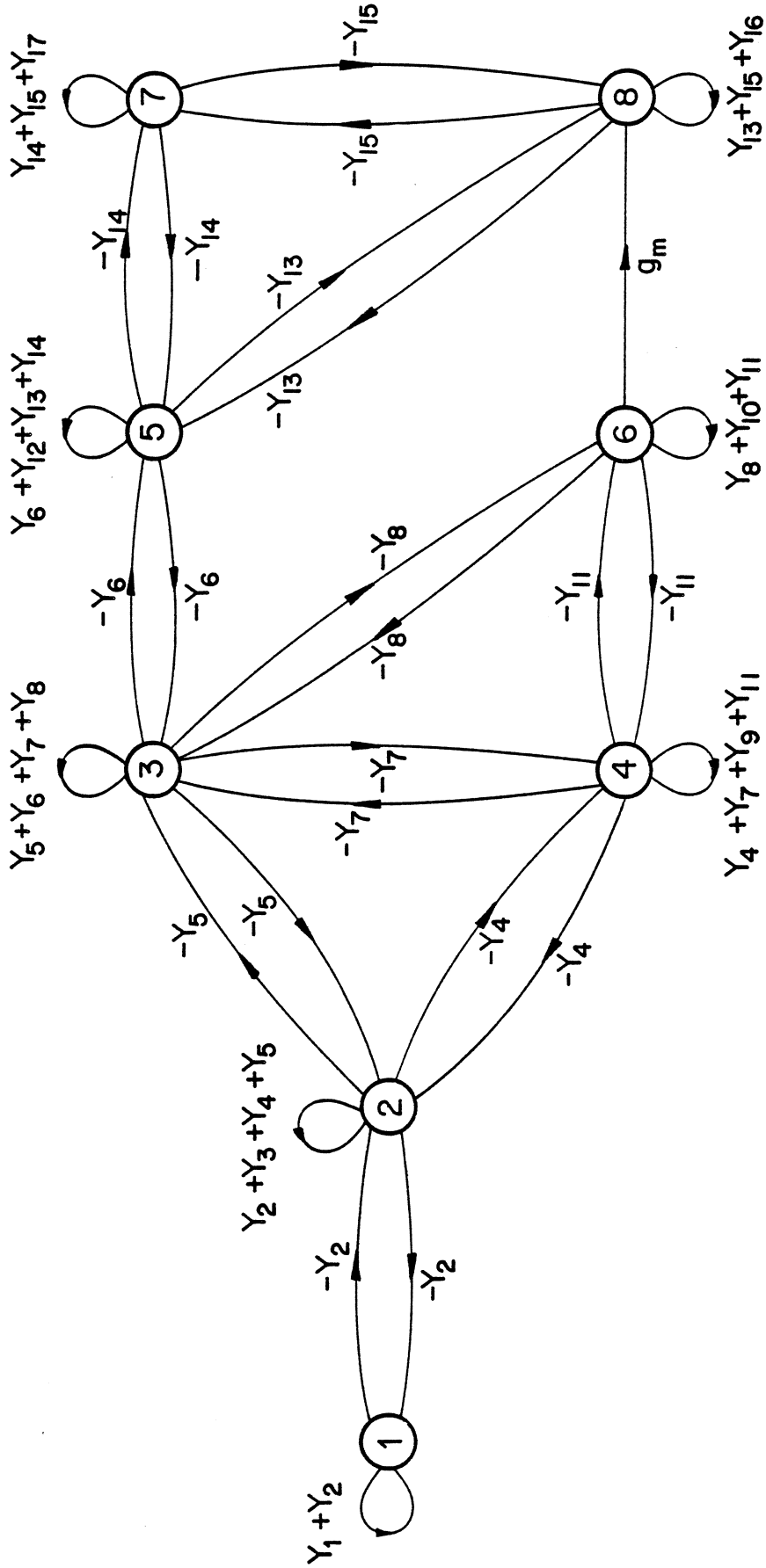


Fig. 7 (b)

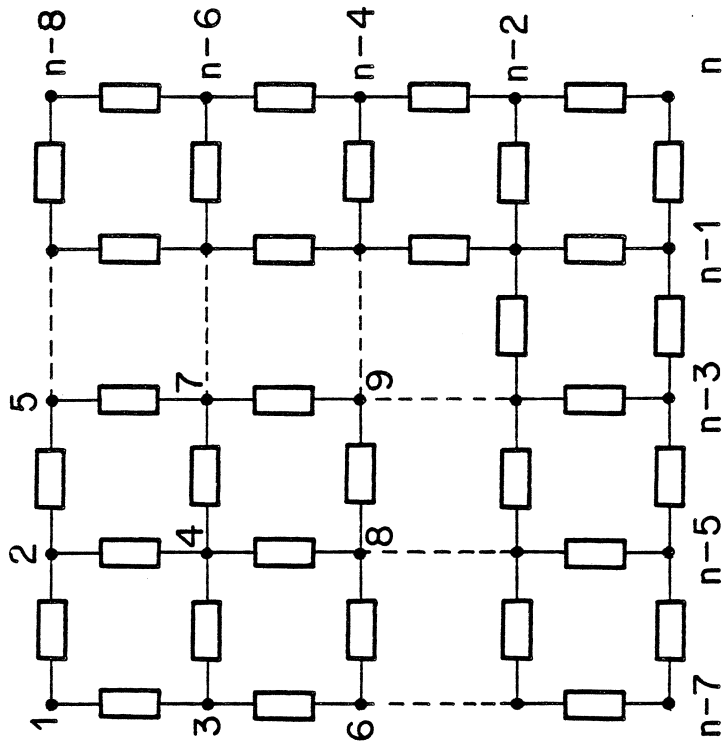


Fig. 8

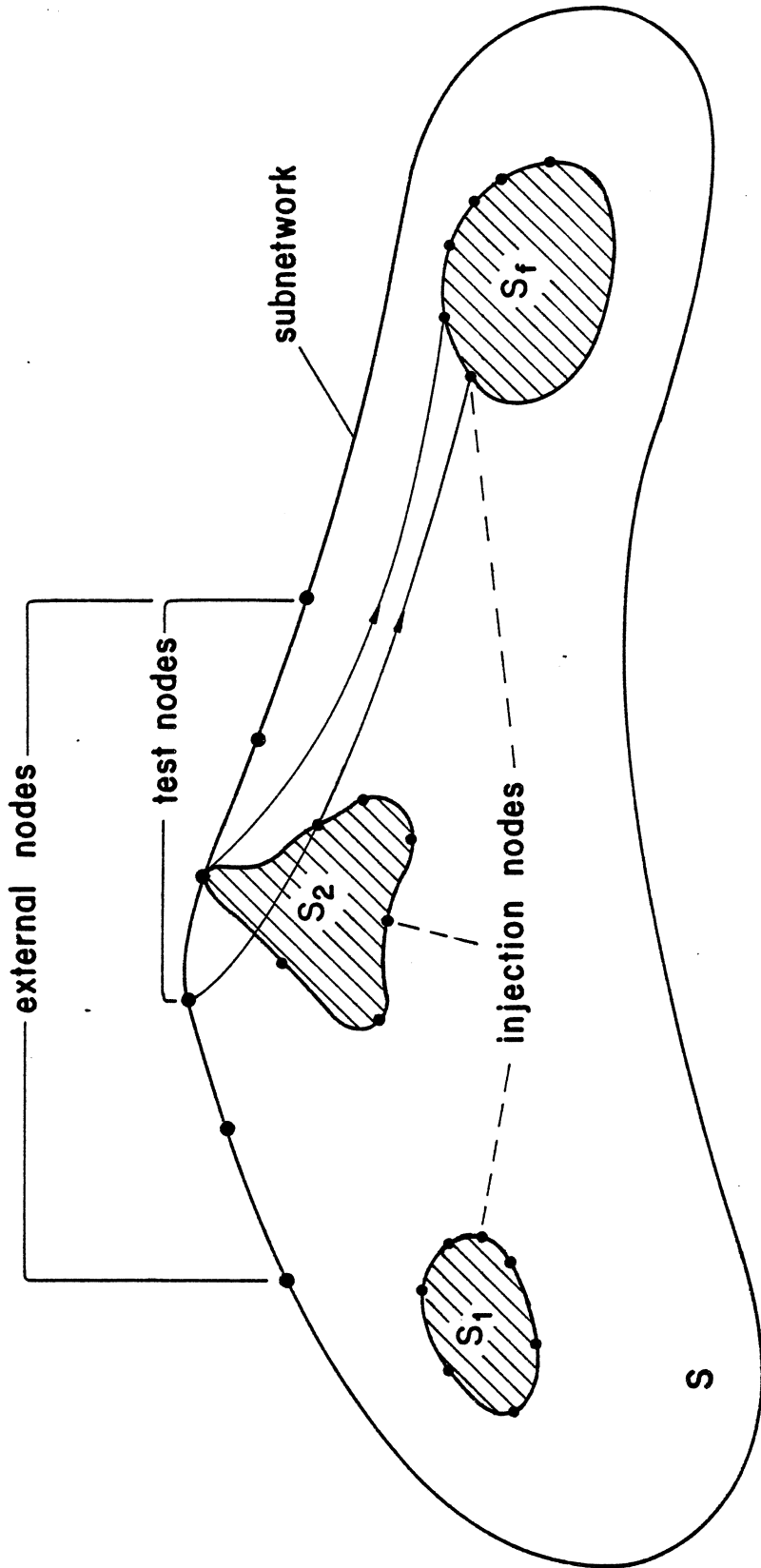


Fig. 9

SOC-305

EVALUATION OF FAULTY ELEMENTS WITHIN LINEAR SUBNETWORKS

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Key Words: Fault location, analog circuit analysis, hierarchical decomposition, parameter identification, network decomposition

Abstract: This paper presents the theoretical background for designing tests which are topologically sufficient for identification of faulty parameter values in linear subnetworks. Nodal voltages are assumed to be obtainable either by measurements or, indirectly, as a result of a nodal fault analysis. A formulation of nodal fault analysis for subnetworks is presented. It is shown how this approach can be used to evaluate faulty elements within inaccessible faulty subnetworks. The objective of this work is the reduction of the number of required current excitations and, thereby, the number of voltage measurements. Coates flow-graph representation of a network is used.

Description:

Related Work: SOC-233, SOC-235, SOC-236, SOC-244, SOC-251, SOC-259, SOC-266, SOC-267, SOC-268, SOC-269, SOC-271, SOC-285.

Price: \$10.00.

